# MULTI-MODE STABILIZATION OF TORSIONAL OSCILLATIONS IN SINGLE AND MULTI-MACHINE SYSTEMS USING EXCITATION CONTROL

by

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ABSTRACT

Subsynchronous Resonance (SSR) phenomena in a thermal-electric power system with series-capacitor-compensated transmission lines may cause damaging torsional oscillations in the shaft of the turbine-generator. This thesis deals with a wide-range multi-mode stabilization of single-machine and multi-machine SSR systems using output feedback excitation control. Chapter 1 summarizes the SSR countermeasures to date. Chapter 2 presents a unified electro-mechanical model for SSR studies, illustrates the torsional interaction between the electrical and mechanical systems, and demonstrates that multi-mass representation of the turbine-generator must be used for SSR For the control design, a reduced order model is desirable. For studies. the model reduction, modal analysis is applied to identify the excitable torsional modes, and a mass-spring equivalencing technique to retain only the unstable modes is developed in Chapter 3. Using the reduced order onemachine models, linear optimal excitation controls are designed in Chapter 4. The controls are further simplified by examining the eigenvalue sensitivity, and the results are tested on the linear and nonlinear full models. In Chapter 5, the stabilization technique is further extended and applied to a two-machine system and a three-machine system. The stabilizers still can be designed one machine at a time using a one-machine equivalent for each machine by retaining only the path with the strongest interacting current and the critical electrical resonance frequency as seen by the machine. To coordinate all controllers for the entire system, an iterative process is The controllers thus designed are tested on linear and nonlinear developed. full models. From both eigenvalue analysis and nonlinear dynamic performance tests of the one-machine, two-machine, and three-machine systems, a conclusion is drawn in Chapter 6 that the excitation controls thus designed by the

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methods developed in this thesis can effectively stabilize single-machine and multi-machine SSR systems over a wide range of capacitor compensation.

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## NOMENCLATURE

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General	
Α	system matrix
В	control matrix
u	control vector
x	state vector of the system A
Q	symmetric positive semi-definite weighting matrix
R	symmetric positive definite weighting matrix
К	Riccati matrix
M	composite matrix as defined in (4.7)
Λ	eigenvalue matrix of M
A <sub>c</sub>	system matrix of the controlled system
F	feedback matrix as defined in (4.21)
λ	eigenvalue of the system
X,V	eigenvector matrix of $A_{c}$ and transpose of $A_{c}$ respectively
• x	time derivative of x
9	partial differential operator
S	Laplace tramsform operator
Δ	prefix denoting a linearized variable
0	subscript denoting initial condition
T	superscript denoting transpose
-1	superscript denoting inverse
j	complex operator $\sqrt{-1}$

## Mass-spring system

M	inertia constant
K	shaft stiffness constant
D	damping coefficient
θ	rotor angular displacement in radian
ω	rotor speed in per unit
ω	synchronous speed which is one per unit
ω	base speed or 377 radian/second
q	modal angular displacement
Qmode	eigenvector matrix of the mass-spring system
Synchronous	Machine
I	instantaneous value of current
v	instantaneous value of voltage
Ψ	flux-linkage
R	resistance
X	reactance
δ	torque angle
Te	electric torque
I <sub>t</sub>	terminal current
v <sub>t</sub>	terminal voltage
P <sub>e</sub> + jQ <sub>e</sub>	generator output power
d,q	subscript denoting direct- and quadrature-axis stator quantities
f	subscript denoting field circuit quantities
kd,kq	subscripts denoting direct- and quadrature-axis damper quantities
a	subscript denoting armature phase quantities
i (=1,2,3)	subscript denoting to which machine the quantity belongs

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## Transmission line

X <sub>ti</sub> ,R <sub>ti</sub>	reactance and resistance of the i-th transformer
X <sub>Li</sub> ,R <sub>Li</sub>	reactance and resistance of the i-th transmission line
X <sub>ci</sub>	reactance of the capacitor in the i-th transmission line
с	subscript denoting quantities associate with capacitor

## Exciter and Voltage Regulator

K <sub>Ai</sub>	gain of the i-th regulator
<sup>T</sup> Ai	time constant of the i-th regulator
T <sub>EI</sub>	time constant of the i-th exciter
Vref	reference voltage

## Governor and Steam Turbine System

K g	actuator gain
<sup>T</sup> 1, <sup>T</sup> 2	actuator time constants
T <sub>3</sub>	servomotor time constant
a	actuator signal
P <sub>GV</sub>	power at gate outlet
тсн	steam chest time constant
TRH	reheater time constant
T <sub>CO</sub>	cross-over time constant
F <sub>HP</sub>	high pressure turbine power fraction
FIP	medium pressure turbine power fraction
F <sub>LPA</sub>	low pressure turbine A power fraction
FLPB	low pressure turbine B power fraction
T <sub>HP</sub>	high pressure turbine torque
T <sub>IP</sub>	medium pressure torque
T <sub>LPA</sub> , T <sub>LPB</sub>	low pressure turbine torque

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#### 1.1 Subsynchronous Resonance

The application of series capacitors to increase the power transfer capability of the transmission system is the best alternative to cope with the ever-increasing electric power damand, the unavaibility of generation sites to build thermal electric power plant at heavy load centers, and the difficulties in obtaining the right of way to build new transmission lines due to envirnomental and economical considerations. However, the series-capacitor-compensated transmission line will cause electrical resonance at certain frequencies which may excite the mechanical mode oscillations of the steam turbine and generator mass-spring system resulting in shaft damage and other detrimental effect to the power The term "Subsynchronous Resonance (SSR)" has been used to desigsystem. nate the oscillating phenomenon of the electrical and mechanical variables associated with turbine-generators connected to transmission systems with series capacitor compensation. Typical example of the damaging effect due to SSR is illustrated in Figure 1.1 ; there were two shaft failures at Mohave generating station in 1970 and 1971 [1].

Despite the hazards of SSR, utilities still favor the use of series capacitors to increase the power transfer capability. In order to overcome the problems caused by SSR, extensive effort has been made in analyzing the two shaft failures. Problems are identified as the induction generator effect, torsional interaction, and transient torques [2].

SSR phenomenon may occur in two different forms : the steady state SSR which is the result of induction generator effect and torsional



(a)



(b)

Figure 1.1 Photographs of

(a) Damaged collector, Mohave No. 1

(b) Cross section from damaged shaft.

\* Pictures are taken from reference [3].

interaction, and the transient SSR which involves the transient torques on segments of the turbine-generator shaft caused by fault or switching operation in the electrical system.

#### 1.2 Countermeasures to SSR

In the past decade, many countermeasures to SSR problems are proposed. Some significant ones are as follows:

1) Static blocking filter [2,4]

High quality factor blocking filters are inserted per-phase in between the high voltage winding of the step-up transformer and the neutral to block electrical resonance currents at the critical frequencies corresponding to the torsional modes of the turbine-generator mass-spring system. When perfectly tuned, the filters will block the currents at critical frequencies completely. But the filters may be detuned due to the change in system frequency after a disturbance or the change in ambient temperature which affects the filter's parametric values. A large space is required to install the filters and the basic insulation level of the step-up transformer must be increased; it is an expensive device.

2) Dynamic filter [ 5 ]

An active device which generates a voltage in series with the generator to nullify the subsynchronous voltage generated by any oscillating motion of the rotor, thereby preventing self excitation due to the torsional interaction. It is unaffected by the system frequency and the number of series capacitors in service. But it is quite costly because of the complex control system and the requirement of an isolated power

source.

3) Dynamic stabilizer [6,7]

The device consists of thyristor controlled shunt reactors connected to the synchronous machine terminal. Control of SSR is achieved by modulating the thyristor switch firing angles to control the reactive power consumed by the reactors.

4) Bypassing the series capacitor

a) The series capacitor is bypassed with the aid of a dual gap which flashes at a lower current level to limit the transient torque build-up, and the gap is reset each time to a higher level allowing a current decay to the level for successful reinsertion of the capacitor. The dual gap flashing scheme can reduce the transient torque significantly [2].

b) In another scheme, called the NGH-SSR scheme [8,9], a thyristor pair in series with a resistor is inserted across the capacitor. By firing the thyristors by some control scheme, the capacitor's charges are dissipated through the resistor. However, this device cannot be used to stabilize the SSR system over a wide range of capacitor compensation [8].

5) Supplemental excitation control

By this method, signal obtained from a properly designed controller is used to modulate the output of the excitation in response to the torsional oscillations of the turbine-generator, and hence provides adequate damping to the SSR system.

Although the shaft failure incidents are ten years old, many ongoing researches are still focused on SSR, indicating that power engineers are still searching for more effective and less expensive means to overcome the problem. Of all the proposed countermeasures of SSR, the supplemental excitation control seems to be the least expensive means which needs further investigation.

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## 1.3 Previous Works of Excitation Control of SSR

Excitation control of SSR involves the use of a control signal to modulate the output of the excitation system to enhance the damping of torsional modes of the generator mass-spring system. Many control schemes have been proposed [ 10-18 ], but it is very desirable to have a controller which can stabilize multi-mode SSR oscillations in a multi-machine power system over a wide range of capacitor compensation and operating conditions. The major approaches of the controller design are two:

### 1) Transfer function approach [ 10-16 ]

A linearized model is used for the controller design and it is mainly based upon the phase compensation concept; a transfer function representing the controller is assumed and the parameters of the controller are chosen such that it can stabilize the system. Since the generator mass-spring system representation plays a very important role in SSR studies, the control design based on a one lumped-mass generator model can only suppress the electrical resonance phenonmena [10-13], and cannot reduce the torsional interaction between the electrical and mechanical systems [12].

Unified electro-mechanical system model is also used in excitation control design [14-16], but the effectiveness of the controllers are verified on the linearized model [14], and neglecting the exciter voltage ceiling limits [15,16]. There is no evidence that those controllers can stabilize the SSR system over a wide range of operating conditions and capacitor compensation.

#### 2) Linear Optimal Excitation Control

For the control design, a unified electrical and mechanical model is developed [17]. A linearized model is used for the control design, based upon modern control laws, and using a linear combination of feedback signals which collectively ensures proper damping to all torsional modes of the system [17,18].

Although application of linear optimal excitation control (LOEC) to power System dynamic stability control is well documented [19-21], it is relatively new in using LOEC for SSR stabilization. To apply LOEC to SSR problem, Yu, Wvong, and Tse had shown as the first step the feasibility of linear stabilization of SSR which is not optimal [17]. The work was continued by Yan, Wvong, and Yu [18] to develop LOEC of SSR. The control was tested on both linear and nonlinear full models. The results indicated that the LOEC can effectively stabilize the SSR system over a wide range of capacitor compensation and operating conditions.

But the design still requires improvement, especially in two aspects

- 1) The order of the model in [ 17,18 ] shall be further reduced for the controller design and the controller designed must be simplified to the extent that only a minimum number of measurable output feedback signals are required to implement the controller. Techniques of further reduction of the model and simplification of the final control must therefore be developed.
- 2) The excitation control of SSR must be applicable not only to the one-machine system but also to a multi-machine system. There is dynamic interaction between machines in a multi-machine SSR system, which may be divided into two categories:

- a) The dynamic interaction of the low frequency oscillations between machines [ 22 ].
- b) The interaction between the torsional modes of the mass-spring system of different machines [ 23 ].

### 1.4 Scope of the Thesis

This thesis deals with the output feedback linear optimal excitation control of one-machine and multi-machine SSR systems. Chapter two recapitulates all the basic equations of power system models for SSR studies. The torsional interaction effect between the electrical and mechanical systems is illustrated. In Chapter three, modal analysis is applied to the generator mass-spring system and a mass-spring equivalencing technique [24] for model reduction is developed. In Chapter four, procedures of linear optimal excitation control design are given, and the designed controller is implemented to the full model for both eigenvalue analysis and nonlinear dynamic performance tests. In Chapter five, the multi-machine SSR problem is examined, one machine equivalents of the multi-machine system are developed, and the excitation controls designed for a two-machine and a three-machine power system are implemented on the full models for eigenvalue analysis and nonlinear dynamic performance tests. Summary of all the important findings is given and a conclusion is drawn in Chapter six.

2. MODELLING POWER SYSTEMS FOR SSR STUDIES

### 2.1 Introduction

To accurately simulate the transient and dynamic behaviour of a power system, a proper and adequate model must be chosen. In conventional power system stability studies for which the low frequency oscillations (.5 - 2 Hz.) is of the main concern, the generator and turbine shaft stiffness, the amortisseur winding effect, and the armature and network transient may be neglected. However, for SSR studies, the emphasis in modelling system components is different. In order to account for the torsional oscillations of the mechanical mass-spring system and the torsional interaction between electrical and mechanical systems, those factors which are not important in conventional stability studies can no longer be neglected.

In the first part of this chapter, a summary of equations describing the power system for SSR studies is given. The complete model for a one machine system is shown in Figure 2.1, which consists of the turbine generator mass-spring system, the speed governor and the turbine torque [25], the generator and excitation [26], and the capacitor compensated transmission line. In the second part of this chapter, the complete system state equations are obtained. It is applicable to both single machine and multi-machine systems. Towards the end of this chapter, the effect of torsional interaction between mechanical and electrical systems is illustrated.





### 2.2 The Mechanical System

Consider the torques of the steam turbines first, they may be written as

$$\dot{T}_{HP} = \frac{F_{HP}}{T_{CH}} P_{GV} - \frac{1}{T_{CH}} T_{HP}$$
(2.1)

$$\dot{T}_{IP} = \frac{F_{IP}}{F_{HP} T_{RH}} T_{HP} - \frac{1}{T_{RH}} T_{IP}$$
 (2.2)

$$\dot{T}_{LPA} = \frac{F_{LPA}}{F_{IP} T_{CO}} T_{IP} - \frac{1}{T_{CO}} T_{LPA}$$
 (2.3)

$$T_{LPB} = \frac{T_{LPB}}{T_{LPA}} T_{LPA}$$

where

P <sub>GV</sub>	:	power at gate outlet
<sup>Т</sup> СН	:	steam chest time constant
T <sub>RH</sub>	:	reheater time constant
<sup>т</sup> со	:	cross-over time constant
F <sub>HP</sub>	:	high pressure turbine power fraction
FIP	:	medium pressure turbine power fraction
F <sub>LPA</sub>	:	low pressure turbine A power fraction
$F_{LPB}$	:	low pressure turbine B power fraction
т <sub>нр</sub>	:	high pressure turbine torque
TIP	:	medium pressure turbine torque
$T_{LPA}$	:	low pressure turbine A torque
T <sub>LPB</sub>	:	low pressure turbine B torque

Consider the mass-spring system next. Assume that there is one high pressure turbine (HP), one medium pressure turbine (IP), two low pressure turbines (LPA,LPB), one generator (Gen), and one exciter (Ex),all

5 (2.4)

on one shaft as shown in Figure 2.1. Although a more accurate mass-spring model is available by modelling shaft and masses in finite sections [34], a linear mass-spring model is recommended by an IEEE committee report for SSR analysis [35]. According to [35], it is assumed that

- (a) There are six lumped masses, each with its inertia constant.
- (b) The shaft between any two masses behaves like a linear torsional spring with negligible mass.
- (c) There is mechanical damping to each rotating mass, although it is very difficult to determine [ 36 ].

Figure 2.2 illustrates the various torsional forces experienced by the i<sup>th</sup> element on the mass-spring system. A positive torsional torque  $K_{i,i+1}(\theta_{i+1} - \theta_i)$  on the right; a negative torque  $-K_{i-1,i}(\theta_i - \theta_{i-1})$  on the left; and an external torque  $T_i$ , a positive accelerating torque  $M_i \omega_i$ , and a negative damping torque  $-D_i \omega_i$  on the mass itself. A general equation of motion of the i<sup>th</sup> rotor is as follows

$$M_{i}\omega_{i} = T_{i} - D_{i}\omega_{i} + K_{i,i+1}(\theta_{i+1} - \theta_{i}) - K_{i-1,i}(\theta_{i} - \theta_{i-1})$$
(2.5)

where

M<sub>i</sub> : inertia constant of the i<sup>th</sup> rotational mass
θ<sub>i</sub> : angular displacement of the i<sup>th</sup> rotational mass
K<sub>i,i+i</sub> : torsional stiffness constant of the shaft between i<sup>th</sup> and i+1<sup>th</sup> rotational mass



Figure 2.2 Modelling of the mass-spring system in the vicinity of the i rotational mass

Apply	ing equation (2.5) to the six mass_spring system as s	hown in
Figure 2.1, twelve	e differential equations are obtained :	
High pressure turbine	$\dot{\omega}_{1} = \frac{K_{12}}{M_{1}} \theta_{2} - \frac{K_{12}}{M_{1}} \theta_{1} - \frac{D_{1}}{M_{1}} \omega_{1} + \frac{T_{HP}}{M_{1}}$	(2.6)
	$\dot{\theta}_1 = \omega_b (\omega_1 - \omega_o)$	(2.7)
Medium pressure turbine	$\dot{\omega}_{2} = \frac{K_{12}}{M_{2}} \theta_{1} - (\frac{K_{12} + K_{23}}{M_{2}}) \theta_{2} + \frac{K_{23}}{M_{2}} \theta_{3} - \frac{D_{2}}{M_{2}} \omega_{2} + \frac{T_{IP}}{M_{2}}$	(2.8)
." 	$\dot{\theta}_2 = \omega_b (\omega_2 - \omega_o)$	(2.9)
Low pressure turbine A	$\omega_3 = \frac{K_{23}}{M_3} \theta_2 - (\frac{K_{23} + K_{34}}{M_3}) \theta_3 + \frac{K_{34}}{M_3} \theta_4 - \frac{D_3}{M_3} \omega_3 + \frac{T_{LPA}}{M_3}$	(2.10)
	$\theta_3 = \omega_b (\omega_3 - \omega_o)$	(2.11)
Low pressure turbine B	$\omega_4 = \frac{K_{34}}{M_4} \theta_3 - (\frac{K_{34} + K_{45}}{M_4}) \theta_4 + \frac{K_{45}}{M_4} \theta_5 - \frac{D_4}{M_4} \omega_4 + \frac{T_{LPB}}{M_4}$	(2.12)
· · · · · · · · · · · · · · · · · · ·	$\theta_4 = \omega_b (\omega_4 - \omega_o)$	(2.13)
Generator	$\dot{\omega} = \frac{K_{45}}{M_5} \theta_4 - (\frac{K_{45} + K_{56}}{M_5}) \delta + \frac{K_{56}}{M_5} \theta_6 - \frac{D_5}{M_5} \omega - \frac{T_e}{M_5}$	(2.14)
	$\delta = \omega_{\rm b} (\omega - \omega_{\rm o})$	<b>(</b> 2.15)
Exciter	$\omega_6 = \frac{K_{56}}{M_6} \delta - \frac{K_{56}}{M_6} \theta_6 - \frac{D_6}{M_6} \omega_6$	(2.16)
· ·	$\theta_6 = \omega_b (\omega_6 - \omega_o)$	<b>(</b> 2.17)
where & : elect	trical angular displacement in electrical radian whic	h is
equal	to the mechanical radian for a two-pole machine.	
ω <sub>-</sub> : speed	d of the i rotor in per unit.	
ب سر: synch	pronous speed which is one per unit.	
ω <sub>L</sub> : base	speed or 377 radian/second	· _
θ:mecha	anical angular displacement in radian.	

Te: electric torque across the air gap in per unit.

## Consider a speed governor next. The state equations may be

written

$$a = \frac{K_g}{T_1} (\omega_{ref} - \omega) - \frac{1}{T_1} a \qquad (2.18)$$

$$\dot{P}_{GV} = \frac{1}{T_3} a - \frac{1}{T_3} (P_{GV} - P_o)$$
 (2.19)

 $\dot{P}_{GV_{min}} \leq \dot{P}_{GV} \leq \dot{P}_{GV_{max}}$ 

where

 ${}^{K}g$  : actuator gain  ${}^{T}1$  : actuator time constant  ${}^{T}3$  : servomotor time constant a : actuator signal  $\omega$  : generator speed  ${}^{\psi}ref$ : reference speed  ${}^{P}o$  : initial power reference

### 2.3 The Electrical System

A transmission system between any two buses is shown in Figure 2.3, where  $R_t$  is the transformer resistance,  $X_t$  is its reactance,  $R_L$  the transmission line resistance,  $X_L$  the line reactance, and  $X_c$  the capacitor reactance. For a one machine infinite system,  $V_t$  denotes the generator terminal voltage,  $V_o$  the infinite bus voltage, and  $V_c$  the voltage across the capacitor.



Figure 2.3

The transmission system

The general voltage equation becomes

$$\begin{bmatrix} V_{t} \end{bmatrix}_{phase} = \begin{bmatrix} R \end{bmatrix} \begin{bmatrix} I_{t} \end{bmatrix}_{phase} + \begin{bmatrix} L \end{bmatrix} \frac{d}{dt} \begin{bmatrix} I_{t} \end{bmatrix}_{phase} + \begin{bmatrix} V_{c} \end{bmatrix}_{phase}$$

$$+ \begin{bmatrix} V_{o} \end{bmatrix}_{phase}$$
(2.20)

In Park's coordinates and for a balanced three-phase operation the d and q components of the generator terminal voltage become

$$V_{d} = \left(\frac{X_{t} + X_{L}}{\omega_{b}}\right) I_{d} - \left(X_{t} + X_{L}\right) I_{q} + \left(R_{t} + R_{L}\right) I_{d}$$

$$+ V_{cd} + V_{od} \qquad (2.21)$$

$$V_{q} = \left(\frac{X_{t} + X_{L}}{\omega_{b}}\right) I_{q} + \left(X_{t} + X_{L}\right) I_{d} + \left(R_{t} + R_{L}\right) I_{q}$$

$$+ V_{cq} + V_{oq} \qquad (2.22)$$

and the two infinite bus voltage components are

$$V_{od} = V_{o} \sin \delta_{o}$$

$$V_{oq} = V_{o} \cos \delta_{o}$$
(2.23)

where  $V_0$  is the magnitude of the infinite bus voltage, and  $\delta_0$  is the angle between machine terminal and the infinite bus. For a multi-machine system,  $V_0$  is not the infinite bus voltage anymore.

The capacitor voltages in the d-q coordinate become

$$\frac{V_{cd}}{\omega_b} = V_{cq} + \omega X_c I_d$$

(2.24)

(2.25)

Equations for exciter and voltage regulator are

$$\dot{E}_{fD} = \frac{1}{T_E} V_R - \frac{1}{T_E} E_{fD}$$
 (2.26)

$$V_{R} = \frac{K_{A}}{T_{A}} (V_{ref} - V_{t} + U_{E})$$
 (2.27)

$$V_{t} = \sqrt{V_{d}^{2} + V_{q}^{2}}$$

$$V_{R_{min}} \leq V_{R} \leq V_{R_{max}}$$
(2.28)

Previous work [ 18 ] found that five winding generator model with one damper winding on each of the d,q-axis is sufficient. The voltage equations of synchronous generator are [37,38]

$$V_{d} = \frac{\Psi_{d}}{\omega_{b}} - \omega \Psi_{q} - R_{a} I_{d}$$
 (2.29)

$$V_{q} = \frac{\Psi}{\omega_{b}} + \omega \Psi_{d} - R_{a} I_{q}$$
(2.30)

$$V_{f} = \frac{\Psi_{f}}{\omega_{b}} + R_{f} I_{f}$$
(2.31)

$$0 = \frac{\Psi_{kd}}{\omega_b} + R_{kd} I_{kd}$$
 (2.32)

$$0 = \frac{\Psi_{kq}}{\omega_b} + R_{kq} I_{kq}$$
(2.33)

and the electric torque equation in per unit is

$$T_e = \Psi_d I_q - \Psi_q I_d$$
 (2.34)

where the flux linkage

$$\begin{pmatrix} \Psi_{d} \\ \Psi_{q} \\ \Psi_{q} \\ \Psi_{f} \\ \Psi_{f} \\ \Psi_{kd} \\ \Psi_{kq} \end{pmatrix} = \begin{pmatrix} -X_{d} & X_{md} & X_{md} \\ -X_{q} & X_{md} & X_{mq} \\ -X_{md} & X_{f} & X_{md} \\ -X_{md} & X_{md} & X_{kd} \\ -X_{mq} & X_{kq} \end{pmatrix} \begin{pmatrix} I_{d} \\ I_{q} \\ I_{f} \\ I_{kd} \\ I_{kd} \\ I_{kq} \end{pmatrix}$$
(2.35)

## 2.4 Complete System Equations for Single and Multi-machine Systems

The original nonlinear equations are used for time domain simulation , and the equations are linearized for eigenvalue analysis. The linearized electrical system equations may be written in the matrix form as follows

$$[B][X_{II}] = [C][X_{I}] + [D][X_{II}]$$
(2.36)

Hence

$$[X_{II}] = [A_{II,I}] [X_{I}] + [A_{II,II}] [X_{II}] (2.37)$$

where

$$[A_{II,I}] = [B]^{-1} [C]$$
 (2.38)  
 $[A_{II,II}] = [B]^{-1} [D]$  (2.39)

 $[X_{I,II}]$  and  $[X_{II,I}]$  represent the interaction between the mechanical system and the electrical system which can be combined with the linearized mechinical system equations to give

$$\begin{pmatrix} \mathbf{X}_{\mathbf{I}} \\ \mathbf{X}_{\mathbf{II}} \\ \mathbf{X}_{\mathbf{II}} \end{pmatrix} = \begin{pmatrix} \mathbf{A}_{\mathbf{I},\mathbf{I}} & \mathbf{A}_{\mathbf{I},\mathbf{II}} \\ \mathbf{A}_{\mathbf{II},\mathbf{I}} & \mathbf{A}_{\mathbf{II},\mathbf{II}} \end{pmatrix} \begin{pmatrix} \mathbf{X}_{\mathbf{I}} \\ \mathbf{X}_{\mathbf{II}} \end{pmatrix}$$
(2.40)

where

$$[X_{I}] = [\Delta \omega_{1}, \Delta \theta_{1}, \Delta \omega_{2}, \Delta \theta_{2}, \Delta \omega_{3}, \Delta \theta_{3}, \Delta \omega_{4}, \Delta \theta_{4}, \Delta \omega, \Delta \delta, \Delta \omega_{6}, \Delta \theta_{6}, \Delta \theta_{6}$$

$$[X_{II}] = [\Delta I_{d}, \Delta I_{q}, \Delta I_{f}, \Delta I_{kd}, \Delta I_{kq}, \Delta V_{cd}, \Delta V_{cq}, \Delta V_{R}, \Delta E_{fd}]^{T}$$
(2.42)

Equation (2.40) may also be written compactly as

$$[X] = [A] [X]$$
 (2.43)

For a multi-machine system where more than one machine is involved, the individual machine coordinate (d,q) may be referred to a common reference frame (D,Q) as follows:



Figure 2.4 Component of I in dq and DQ coordinates

where

$$\begin{pmatrix} \mathbf{I}_{\mathrm{D}} \\ \mathbf{I}_{\mathrm{Q}} \end{pmatrix} = \begin{pmatrix} \cos\theta & -\sin\theta \\ & & \\ \sin\theta & \cos\theta \end{pmatrix} \begin{pmatrix} \mathbf{I}_{\mathrm{d}} \\ & \\ \mathbf{I}_{\mathrm{q}} \end{pmatrix}$$
 (2.44)

Linearized,

$$\begin{pmatrix} \Delta \mathbf{I}_{\mathbf{D}} \\ \Delta \mathbf{I}_{\mathbf{Q}} \end{pmatrix} = \begin{pmatrix} \cos\theta_{0} & -\sin\theta_{0} \\ \sin\theta_{0} & \cos\theta_{0} \end{pmatrix} \begin{pmatrix} \Delta \mathbf{I}_{d} \\ \Delta \mathbf{I}_{q} \end{pmatrix} + \begin{pmatrix} -(\mathbf{I}_{do}\sin\theta_{0} + \mathbf{I}_{qo}\cos\theta_{0}) \\ (\mathbf{I}_{do}\cos\theta_{0} - \mathbf{I}_{qo}\sin\theta_{0}) \end{pmatrix} \Delta \theta \quad (2.45)$$

Therefore

$$\begin{pmatrix} \Delta I_{D} \\ \vdots \\ \Delta I_{Q} \end{pmatrix} = \begin{pmatrix} \cos\theta_{0} & -\sin\theta_{0} \\ \sin\theta_{0} & \cos\theta_{0} \end{pmatrix} \begin{pmatrix} \Delta I_{d} \\ \vdots \\ \Delta I_{q} \end{pmatrix} + \begin{pmatrix} -(I_{do}\sin\theta_{0} + I_{qo}\cos\theta_{0}) \\ (I_{do}\cos\theta_{0} - I_{qo}\sin\theta_{0}) \end{pmatrix} \Delta \theta \quad (2.46)$$

Inversely,

$$\begin{pmatrix} \mathbf{I}_{d} \\ \mathbf{I}_{q} \end{pmatrix} = \begin{pmatrix} \cos\theta & \sin\theta \\ -\sin\theta & \cos\theta \end{pmatrix} \begin{pmatrix} \mathbf{I}_{D} \\ \mathbf{I}_{Q} \end{pmatrix}$$
 (2.47)

Linearized

$$\begin{pmatrix} \Delta \mathbf{I}_{d} \\ \Delta \mathbf{I}_{q} \end{pmatrix} = \begin{pmatrix} \cos\theta_{0} & \sin\theta_{0} \\ -\sin\theta_{0} & \cos\theta_{0} \end{pmatrix} \begin{pmatrix} \Delta \mathbf{I}_{D} \\ \Delta \mathbf{I}_{Q} \end{pmatrix} + \begin{pmatrix} (\mathbf{I}_{Qo}\cos\theta_{0} - \mathbf{I}_{Do}\sin\theta_{0}) \\ -(\mathbf{I}_{Do}\cos\theta_{0} + \mathbf{I}_{Qo}\sin\theta_{0}) \end{pmatrix} \Delta \theta \quad (2.48)$$

Therefore

$$\begin{pmatrix} \Delta \mathbf{I}_{d} \\ \vdots \\ \Delta \mathbf{I}_{q} \end{pmatrix} = \begin{pmatrix} \cos\theta_{0} & \sin\theta_{0} \\ -\sin\theta_{0} & \cos\theta_{0} \end{pmatrix} \begin{pmatrix} \Delta \mathbf{I}_{D} \\ \vdots \\ \Delta \mathbf{I}_{Q} \end{pmatrix} + \begin{pmatrix} (\mathbf{I}_{Qo}\cos\theta_{0} - \mathbf{I}_{Do}\sin\theta_{0}) \\ -(\mathbf{I}_{Do}\cos\theta_{0} + \mathbf{I}_{Qo}\sin\theta_{0}) \end{pmatrix} \Delta \boldsymbol{\theta} \quad (2.49)$$

For the multi-machine formulation, it is convenient to write  

$$X_{I} = [X_{I1}, X_{I2}, X_{I3}, \dots, X_{Ii}]^{T}$$

$$X_{II} = [X_{II1}, X_{II2}, X_{II3}, \dots, X_{III}]^{T}$$
(2.50)

where X<sub>II</sub>, X<sub>III</sub> are the mechanical and electrical state variables of machine i, respectively.

## 2.5 The Torsional Interaction

When an electrical resonance occurs in the electrical system (including the generator stator and transmission line) at a subsynchronous frequency  $f_e$ , it will interact with the rotor and induce a pulsating torque at the frequency of ( 60 -  $f_e$  ), which becomes a forcing torque to the mass-spring system. If the frequency of the oscillating torque equals to a torsional modal frequency  $f_m$ , the electrical resonance and the particular torsional mode will be mutually excited, and a voltage will be induced in stator winding at frequency  $f_e = 60-f_m$ . Thus the torsional interaction looks like a negative resistance to the electrical system and negative damping to the torsional system under these conditions.

To illustrate the effect of torsional interaction, the one machine infinite bus system as shown in Figure 2.1 is tuned so that SSR will occur.

The variation of electric power which is approximately equal to the electric torque in the per unit system, as shown in Figure 2.5, roughly consists of two components: a low frequency oscillation in the range of 1 - 2 Hz., and a higher frequency oscillation. In this particular case, the subsynchronous torque increases with time.

The results of Figure 2.5 also can be synthesized as follows : First let the system be modified by lumping the six torsional masses into one. The power variations for the same system disturbance are shown as Figure 2.6. The responses as shown in Figure 2.5 and 2.6 are the same within 0.5 second but there is no subsynchronous torque component in Figure 2.6 because of the modelling.

A constant negative resistance is then inserted in the transmission line of the above modified system. The system response for the same disturbance is shown in Figure 2.7. Although the mass-spring system model has been simplified, the subsynchronous torque component is substantial. The concept that the torsional system looks like a negative resistance to the electrical system is further verified. Of course, a static negative resistance representation of the torsional interaction is oversimplified in the SSR studies.



Figure 2.5 Power variation of the one-machine infinite bus system with the mass-spring system modelled in detail.





Power variation of the one-machine infinite bus system with the mass-spring system lumped into one mass and a constant negative resistance inserted in the transmission line.

3. MODEL REDUCTION OF A POWER SYSTEM FOR SSR STUDIES

### 3.1 Introduction

The requirement of including the torsional mass-spring system, the generator, and transmission system in one single unified model for SSR studies so that the torsional interaction of the electrical and mechanical systems will be automatically included inevitably results in a very high order system. For instance, the full model of a one-machine infinite-bus system for SSR studies developed in the last chapter is of 26th order. For some SSR studies, especially for the control design, a reduced order model is very desirable.

For an excitation control design, the steam turbine torque and governor equations together with a small time constant of the exciter can be neglected. The order of the model is reduced from 26th to 20th.

However, further order reduction is still necessary. A new technique of obtaining a reduced order equivalent mass-spring system is developed in this chapter by retaining the unstable torsional modes alone without changing their oscillating frequencies. The first step is to determine the relative instability among the unstable modes using modal analysis. The second step is to retain only the unstable torsional modes. Finally, the eigenvalues of the original model and the reduced order models are compared.
#### 3.2 Torsional Resonance and the Unstable Modes

The prevailing techniques in SSR stability studies are either the frequency scanning method together with the torsional interaction equations [27] to determine the stability of the torsional modes one at a time, or apply either eigenvalue analysis or Nyguist criteria [28] to the unified electro-mechanical power system. Without the analysis, no one could confidently predict which mode is more vulnerable to torsional oscillations than others. In this section, a technique to identify the excitable torsional modes, or the mode which is vulnerable to torsional interaction, and to determine the relative instability of the unstable modes is presented. The technique requires no information about the electrical network, and is based on the modal analysis of the torsional mass-spring system alone.

## 3.2.1 Natural Frequencies and Mode Shapes of the Mass-spring System

The natural frequencies and the mode shapes of the turbine generator mass-spring system can be found as follows:

Consider an unforced and undamped mass-spring system as shown on top of Figure 3.1. The system can be described by a set of second order differential equations in matrix form as follows

$$\frac{1}{\omega_{b}} [M] [\theta] + [K] [\theta] = 0$$
(3.1)

where

$$[ \theta ] = \text{diag} [ M_1, M_2, M_3, M_4, M_5, M_6 ]$$
(3.2)  
$$[ \theta ] = [ \theta_1, \theta_2, \theta_3, \theta_4, \theta_5, \theta_6 ]^{\text{T}}$$
(3.3)



Figure 3.1 Mode Shapes of the six mass-spring system

$$[K] = \begin{pmatrix} K_{12} & -K_{12} & 0 & 0 & 0 & 0 \\ -K_{12} & (K_{12} + K_{23}) & -K_{23} & 0 & 0 & 0 \\ 0 & -K_{23} & (K_{23} + K_{34}) & -K_{34} & 0 & 0 \\ 0 & 0 & -K_{34} & (K_{34} + K_{45}) & -K_{45} & 0 \\ 0 & 0 & 0 & -K_{45} & (K_{45} + K_{56}) & -K_{56} \\ 0 & 0 & 0 & 0 & -K_{56} & K_{56} \end{pmatrix}$$
(3.4)

where the subscripts 1,2,3,...,6 correspond to HP, IP, LPA, LPB, Gen, Ex respectively.

Assume that all masses oscillate at a resonant frequency  $\omega_{\mbox{m}}$  such that

$$\theta_{i} = X_{i} \sin(\omega_{m} t + \alpha)$$
  $i = 1, 2, ..., 6$  (3.5)

Substituting equation ( 3.5) into (3.1) gives

$$\left(-\frac{\omega_{\rm m}^2}{\omega_{\rm b}}\right)$$
 [X] - [M]<sup>-1</sup>[K][X] = 0 (3.6)

or

$$[M]^{-1}[K][X] = \lambda_{m}[X]$$
(3.7)

where

$$\lambda_{\rm m} = \omega_{\rm m}^2 / \omega_{\rm b} \tag{3.8}$$

Therefore  $\lambda_m$  can be obtained by solving (3.7) and mode frequency  $\omega_m$  can be calculated from

$$\omega_{\rm m} = \sqrt{\lambda_{\rm m} \omega_{\rm b}} \qquad {\rm m} = 0, 1, \dots, 5 \qquad (3.9)$$

There is an eigenvector  $X_m$  corresponding to each eigenvalue  $\lambda_m$ , which, when normalized with respect to the element with the largest magnitude gives the mode shape of that particular mode. The mode frequencies and

the mode shapes of the mass-spring system are given in Figure 3.1. A mode shape gives the relative displacement of each spring-mass during normal mode vibration ( when one particular mode is excited ), but it gives no information about which mode is more unstable than the others. An alternative method will be introduced to overcome this difficulty in the subsequent sections.

## 3.2.2 Damping and the Resonant Peak

Figure 3.2 shows a torsional mass-spring system with an inertia constant M, a stiffness constant K, a damping coefficient D, and a forcing function  $T_o$  sin $\omega$ t



//Figure 3.2 A torsional mass-spring system

The equation of motion of the system is

$$T = M \ddot{\theta} + D \dot{\theta} + K \theta \qquad (3.10)$$

For a solution  $\theta = X \sin(\omega t - \phi)$ , we have

$$\frac{T \sin \omega t}{X} = (K - M\omega^2) \sin(\omega t - \phi) + D\omega \cos(\omega t - \phi) \quad (3.11)$$

In phasor notation

$$\frac{T_o}{X} = (K - M\omega^2) + j D\omega$$

26

(3.12)

Hence

$$X = \frac{T_o}{(K - M\omega^2) + j D\omega}$$

and the phase angle,  $\phi$  is given by

$$\tan \phi = \frac{D\omega}{(K - M\omega^2)}$$
(3.14)

When the frequency of the forcing function equals the natural frequency of the system, the real part of the denominator of (3.13) vanishes and the equation becomes

$$X = \frac{1}{jD\omega}$$
(3.15)

During resonance, the amplitude X is directly proportional to the applied force and inversely proportional to the system damping. Therefore, a large applied force together with a small system damping will result in a large resonant peak.





Figure 3.3 A damped six mass-spring system with unity sinusoidal forcing torque.

Consider a damped mass-spring system with unity sinusoidal forcing torque sin  $\omega$ t applied on the generator rotor as shown in Figure 3.3. The matrix equation in per unit describing the dynamical behaviour of the system becomes

$$\frac{1}{\omega_{b}} [M] [\theta] + \frac{1}{\omega_{b}} [D] [\theta] + [K] [\theta] = [T]$$
(3.16)

27

(3.13)

where [M] and [K] respectively are the inertia constant and stiffness coefficient matrices as shown in equations (3.2) and (3.4), and

$$[D] = diag [D_1, D_2, D_3, D_4, D_5, D_6]$$
$$[T] = [0 \ 0 \ 0 \ 0 \ \sin \omega t \ 0]^T$$

Let the mechanical angular displacements [ $\theta$ ] be transformed into modal angular displacements [q] by the eigenvector matrix [ $Q_{mode}$ ], from the eigenvalue analysis of the undamped system

$$[\theta] = [Q_{mode}][q]$$
 (3.17)

Substituting (3.17) into (3.16) and premultiplying the whole equation by  $[Q_{mode}]^T$ , we have

$$\frac{1}{\omega_{b}} \begin{bmatrix} M_{mode} \end{bmatrix} \begin{bmatrix} q \end{bmatrix} + \frac{1}{\omega_{b}} \begin{bmatrix} D_{mode} \end{bmatrix} \begin{bmatrix} q \end{bmatrix} + \begin{bmatrix} K_{mode} \end{bmatrix} \begin{bmatrix} q \end{bmatrix}$$

$$= [T_{mode}]$$
 (3.18)

where

$$[M_{mode}] = [Q_{mode}]^{T} [M] [Q_{mode}]$$

$$[K_{mode}] = [Q_{mode}]^{T} [K] [Q_{mode}]$$

$$[D_{mode}] = [Q_{mode}]^{T} [D] [Q_{mode}]$$

$$[T_{mode}] = [Q_{mode}]^{T} [T]$$

$$(3.19)$$

Note that [T<sub>mode</sub>] indicates the contribution of the applied force on each mode of vibration; both [M<sub>mode</sub>] and [K<sub>mode</sub>] are diagonal matrices because [M] and [K] are symmetrical.

Neglecting the off diagonal elements of [ D mode], equation

(3.19) becomes six second order differential equations each of which corresponds to (3.10). Applying (3.15), the calculated magnitude of the resonant peaks for various modes for the unity forcing function sin wt of Figure 3.3 are shown in Table 3.1. Multiplying the modal resonance peaks by its corresponding mode shapes as shown in Figure 3.1, gives the relative magnitude of the 'angular displacement' of each mass at the resonant frequencies. Results are shown in Table 3.2.

Table 2.1	Та	ble	3	.1
-----------	----	-----	---	----

Modal resonance peaks of the massspring system

mode 1	6.47
mode 2	1.00
mode 3	3.075
mode 4	4.56
mode 5	0.035

Table 3.2

Magnitude of the resonant peaks of each mass using approximate modal analysis

-							
		HP	IP	LPA	LPB	Gen	Ex
	mode <u>1</u>	4.97	3.73	2.20	0.72	2.39	6.47
	mode 2	0.125	0.07	0.017	0.05	0.04	1.00
	mode 3	3.075	1.04	0.70	0.29	0.51	0.77
	mode 4	3.94	0.20	2.29	4.56	2.83	1.72
	mode 5	0.027	0.035	0.004	0.0007	0.0002	0.00003

3.2.4 Identification of the Unstable Torsional Modes

Neglecting the off diagonal elements of  $[D_{mode}]$ , equation (3.18) becomes six second order differential equations, each corresponds to one mode of vibration. Excluding mode 0, we shall have

 $\frac{1}{\omega_{b}} \underset{\text{mode}_{i}}{\overset{q}{_{i}}} + \frac{1}{\omega_{b}} \underset{\text{mode}_{i}}{\overset{q}{_{i}}} + \underset{\text{mode}_{i}}{\overset{q}{_{i}}} = \underset{\text{mode}_{i}}{\overset{mode}{_{i}}}, \quad i=1,..,5 \quad (3.20)$ when the unity forcing torque since is applied to the system, it is found that is this particular study

$$[T_{mode}] = [0.373, 0.0374, 0.166, 0.6205, 0.0045]$$

In the mean time

 $[D_{mode}] = [0.22, 0.102, 0.127, 0.253, 0.163]$ 

When these results are compared with those from eigenvalue analysis as shown in Table 3.3

#### Table 3.3

Unstable modes of the SSR System							
Mode	Frequency	Compensation	Eigenvalue				
0	1 – 2 Hz	below 30%					
1	15.7 Hz	above 70%	+1.7178±j102.16				
4	32.3 Hz	at 50%,60%	+0.7094±j203.32				
3	25.5 Hz	at 40%,50%,60%,70%	+0.5595±j161.00				

there is an indication that if

$$(d_{mode} - t_{mode}) < 0$$

the particular torsional mode is vulnerable to instability in this particular study.

(3.21)

#### 3.2.5 Modal Resonance Peaks and Unstable Modes

The modal resonance peaks in Table 3.1 indicate the relative instability among the torsional modes. A large modal resonance peak means a large amplitude of vibration ,resulting in large negative damping due to the torsional interaction, and vice versa.

In additional to the six mass-spring system, several other systems are investigated using both eigenvalue analysis of the unified electro-mechanical model and modal resonance peak analysis of the massspring system. All results suggested that the modal resonance peaks can be used as an index to determine the relative instability among the unstable modes. In this particular case, mode 1 will be the most unstable one followed by mode 4 and mode 3 in that order.

#### 3.2.6 Other uses of modal analysis in SSR Studies

Stability of a SSR system also depends upon the conditions of the electrical system. However, the modal analysis is a very useful first step to identify the excitable modes so that the range of frequency scanning can be narrowed, and the number of equations can be reduced. With the excitable modes identified, one can also construct equivalent massspring system by retaining only the unstable modes alone, which will be presented in the next section.

#### 3.3 Equivalent Mass-spring system

As shown in Table 3.3, there are only three torsional modes in the study system. A low order equivalent mass-spring system may be obtained by retaining only the unstable modes as follow:

1) Mode 2 in Figure 3.6a is roughly equal to the exciter mass swings against the rest of the mass-spring system which can be eliminated according to the ongoing analysis. Combining the generator (Gen) and the exciter (Ex) masses together, eliminating mode 2, and keeping all other natural frequencies by adjusting the stiffness constant  $K_{45}$  between the low pressure turbine B (LPB) and the generator (Gen), result in an equivalent five mass-spring system.

2) The same procedure is applied to the high pressure turbine (HP) and the medium pressure turbine (IP). Combining the two masses and eliminating mode 5 by adjusting the stiffness constant  $K_{23}$  between the medium pressure turbine (IP) and the low pressure turbine A (LPA) result in an equivalent four mass-spring system.

3) Finally, the procedure is applied to the two low pressure turbines (LPA) and (LPB). Combining the two masses and elminating mode 1 by adjusting both stiffness constant  $K_{23}$  and  $K_{45}$  but one at a time, result in an equivalent three mass-spring system.

3.3.1 Mode Frequencies and the Adjustment of the Stiffness Constant

Mode frequencies of the mass-spring system depend upon the inertia constants and the stiffness constants. To retain other natural frequencies after combining the two neighbouring masses into one, the stiffness constants must be varied, which can be found as follows:

1) The initial eigenvalues are found from the characteristic equation of the six mass-spring system, each corresponds to a second order equation, and together they may be written in matrix form as

$$\lambda [I] - [M]^{-1} [K] | = 0$$
 (3.22)

where

[ M ] is the inertia constant matrix of appropriate dimension
[ K ] is the spring coefficient matrix

[I] is the identity matrix

$$\lambda = \omega^2/\omega_b$$
,  $\omega_b = 377 \text{ rad/sec.}$ 

2) To find a reduced order mass-spring system, one of the stiffness constant, say  $K_{ij}$ , is treated as an unknown. By substituting the known  $\lambda$ 's already obtained from step (1) one at a time, an average  $K_{ij}$  is obtained.

3) The average  $K_{ij}$  value is substituted into equation (3.22) of the reduced order system, and eigenvalues are recalculated. The value is then adjusted according to the sensitivity coefficient,  $\partial \omega_i / (\partial K_{ij})$ , until the least square error of the natural frequencies of the equivalent reduced order system is the minimum. The least square error

$$e_{\omega} = \sum_{i=1}^{n} (\omega_{i} - \omega_{i})^{2}$$
 (3.23)

versus the variation of  $K_{45}$  of the equivalent five mass-spring system and  $K_{23}$  of the four mass-spring system are shown in Figure 3.4 and 3.5 respectively.





#### 3.3.2 Mode Shapes of the Original and Equivalent Systems

The eigenvalues of a system are unique, but the eigenvectors are not and can be normalized or multiplied by any non-zero scalar. Although an equivalent reduced order mass-spring system retains all the dominant eigenvalues of the original system, the validity remains to be proven: a reduced order model of the equivalent system shall have almost the same dominant eigenvector of the original system. In addition, the modal displacement of  $M_{Gen}$  shall be close to that of the original system so that the torsional interaction between the electrical and mechanical system can be accurately accounted for.

In this section, the normalized eigenvectors of the original and reduced order mass-spring systems are compared. Excluding mode 0, by which all the masses swing in unison, the eigenvectors of the original system before normalization are

(	-1.0428	-1.5584	-2.2820	+0.5914	-0.6161
	-0.7833	-0.9168	-0.4041	-0.0299	+0.7825
	-0.4595	-0.2128	+0.2713	-0.3442	-0.0886
	+0.1499	+0.5601	+0.1127	+0.6847	+0.0165
	+0.5007	+0.5297	-0.1960	-0.4248	-0.0350
	+1.3420	-14.182	+0.2982	+0.2579	+0.0007

After normalization, they become

	,	•			
	_0.7770	+0.1099	1	+0.864	-0.7870
	-0.5840	+0.0650	+0.3420	-0.0440	1
	-0.3420	+0.0150	-0.2290	-0.5030	-0.1130
	+0.1120	-0.0400	-0.0950	1	+0.0210
	+0.3730	-0.0370	+0.1660	-0.6210	-0.0045
-	1	1	-0.2530	+0.3770	+0.0009

(3.24)

(3.25)

After eliminating mode 2, the eigenvectors of the reduced order five mass equivalent system by retaining mode frequencies 15.94 Hz., 25.46 Hz., 32.28 Hz., and 47.46 Hz., before normalization are

+1.0642	+0.9543	+0.6536	+0.6161
+0.7916	+0.3310	-0.0332	-0.7825
+0.4532	-0.2135	-0.3805	+0.0887
-0.1719	-0.1009	+0.7573	-0.0166
-0.5087	+0.1467	-0.4414	+0.0034

Scaling the modal eigenvectors with respect to the original system before normalization (e.g.,  $\frac{-1.0428}{1.0642}$  x column 1,  $\frac{-1.1810}{0.9543}$  x column 2,  $\frac{0.5914}{0.6536}$  x column 3, 1 x column 4 ) gives

-1.0428	-1.1810	+0.5914	+0.6161	]
-0.7757	-0.4096	-0.0300	-0.7825	
-0.4441	+0.2642	-0.3443	+0.0887	
+0.1684	+0.1249	+0.6852	-0.0166	
+0.4985	-0.1816	-0.3994	+0.0034	]

Normalized with respect to the original system yields

-0.7770	1	+0.8630	-0.7870	
-0.5780	+0.3470	-0.0400	1	
-0.3310	-0.2340	-0.5020	-0.1130	
+0.1250	-0.1060	1	+0.0210	
+0.3710	+0.1540	-0.5830	-0.0045	

After eliminating mode 5, by retaining the mode frequencies 15.82 Hz., 25.49 Hz., and 32.47 Hz., the eigenvectors of the four mass equivalent system before normalization are

36

(3.27)

(3.28)

ſ	-0.9534	-0.8973	+0.3436
-	-0.4477	+0.3383	-0.4378
	+0.1785	+0.1617	+0.7580
	+0.5174	-0.2231	-0.4340

Scaling the eigenvectors with respect to the original system before normalization ( e.g.,  $\frac{0.4595}{0.4477}$  x column 1,  $\frac{0.2713}{0.3383}$  x column 2,  $\frac{0.6847}{0.7580}$  x column 3 ) we have

•	-0.9785	-0.7196	+0.3103	
	-0.4594	+0.2713	-0.3955	
	+0.1832	+0.1297	+0.6847	
	+0.5269	-0.1869	-0.3920	

Normalization yields

ſ	-0.7290	+0.6090	+0.4530	•
	-0.3420	-0.2300	-0.5780	
	+0.1360	-0.1090	1.	
l	+0.3930	+0.1970	-0.5730	

Using the results in (3.25), (3.28), and (3.31), the mode shapes of the original, the five mass equivalent, and the four mass equivalent systems are shown in Figure 3.6a, 3.6b, and 3.6c respectively.

Note that the mode shapes of the equivalent systems are close to those of the original system.

37

(3.29)

(3.30)

(3.31)



Figure 3.6 Mode Shapes of the original and Equivalent Systems

ω 8

## 3.3.3 Eigenvalues of the original and reduced order models

After reducing the order of the original system model from 26th to 20 by neglecting turbine torque and governor equations, etc., the order is further reduced using mass-spring equivalencing technique developed in 3.3, resulting in a 16th or a 14th order model. The 14th order model, however, is valid up to 70% capacitor compensation because mode 1 is not considered in the model. Eigenvalues of the 26th, 16th, and 14th order models over a wide range of capacitor compensation are examined. Typical values are shown in Table 3.4 to 3.6. The 16th order model retains all the dominant eigenvalues over a wide capacitor range, and the 14th order model also retains most of the dominant properties except for very high compensation.

Ta	Ы	е	3	.4

Eigenvalues of various order SSR model at 30% capacitor compensation for  $P_e = 0.9$  p.u. at 0.9 power factor lagging and  $V_t = 1.0$  p.u.

	26th order model	reduced 16th order model	reduced 14th order model
Mechanical modes	-0.1818±j298.18 -0.4938±j203.59 -0.2513±j160.64 -0.6705±j127.02 -0.2810±j99.136 -0.0031±j8.4105	-0.4046±j204.71 -0.2069±j160.05 -0.2289±j99.826 -0.0399±j8.4284	-0.2723±j200.37 -0.4052±j157.16 -0.0229±j8.4469
Turbine and Governor	-0.1418 -4.5826 -3.1056 -4.6721±j0.5722		
Stator and Network	-7.0419±j542.94 -5.5469±j210.33	-7.0444±j542.93 -5.6283±j210.34	-7.0440±j542.93 -5.6596±j210.69
Machine rotor	-8.6469 -31.876 -2.0220	-8.5228 -32.611 -2.2563	-8.5180 -32.627 -2.2569
Exciter and voltage regulator	-499.97 -101.94	-101.57	-101.56

Table 3.J	Та	ble	3.	5	
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Eigenvalues of various order SSR model at 50% coapcitor compensation for  $P_e = 0.9$  p.u. at 0.9 power factor lagging and  $V_t = 1.0$  p.u.

	•		
	26th order model	reduced 16th order model	reduced 14th order model
Mechanical modes	-0.1818±j298.18 +0.1237±j202.88 +0.2603±j161.38 -0.6829±j127.06 -0.3524±j99.345 -0.2327±j9.4692	+0.1033±j204.01 +0.2341±j161.72 -0.3053±j100.05 -0.2566±j9.4905	+0.0167±j200.32 +1.0147±j160.11 -0.2408±j9.5172
Turbine and governor	-0.1418 -3.8415 -3.5122 -4.8239±j0.2945		
Stator and Network	-7.0969±j591.27 -6.1493±j161.74	-7.0987±j591.27 -6.0584±j161.72	-7.0986±j591.27 -6.9236±j159.65
Machine rotor	-8.2324 -32.776 -1.9458	-8.3289 -32.776 -1.9458	-8.2283 -33.475 -2.1670
Exciter and voltage regulator	-499.97 -101.76	-101.76	-101.44

Та	Ъ1	e.	3	•	6

Eigenvalues of various order SSR model at 80% capacitor compensation for  $P_e = 0.9$  p.u. at 0.9 power factor lagging and  $V_t = 1.0$  p.u.

·			
	26th order model	reduced 16th order model	reduced 14th order model
Mechanical modes	-0.1818±j298.18 +0.0134±j202.91 -0.0967±j160.52 -0.5998±j126.95 +1.7178±j102.17 -0.7619±j11.662	+0.0125±j204.04 -0.0496±j159.93 +2.1999±j102.78 -0.7669±j11.682	-0.0409±j200.33 +0.2981±j156.59 -0.7572±j11.737
Turbine and governor	-0.1419 -3.4784±j0.5960 -4.9841±j0.0792		
Stator and Network	-7.1710±j648.08 -6.7654±j103.02	-7.1523±j648.08 -7.1328±j103.06	-7.1523±j648.08 -5.0848±j106.26
Machine rotor	-7.6295 -35.049 -1.7807	-7.5898 -35.594 -1.9481	-7.5873 -35.613 -1.9489
Exciter and voltage regulator	-499.97 -101.43	-101.15	-101.15

#### 4. EXCITATION CONTROL DESIGN

#### 4.1 Introduction

The main objective of the controller design is to stabilize all unstable modes over a wide range of power and capacitor compensation with minimum number of feedback signals.

For the multi-mode stabilization of torsional oscillations, the phase compensation power system stabilizer with single signal input is inadequate for the narrow frequency band sensitivity. It may also have detrimental effects on other torsional modes [17]. Multiple loop lead-lag compensation excitation control has also been designed [14], but remains to be improved.

In this chapter, the state regulator problem of control theory is applied, and the linear optimal excitation control is designed for the multiple torsional mode stabilization. It is a linear combination of many system feedback signals. Instead of the phase compensation, the linear combination of feedback signals according to the control law collectively ensures proper damping for all torsional modes.

In engineering practice, it is also desirable to have the minimum number of feedback signals which can be easily measured. Due to the complexity of the SSR problem, the order of the system model is usually very high and the linear optimal control designed usually requires a large number of feedback signals [17,18]. Suboptimal control algorithms are avaiable in the control literature [29,30], but heavy computation is involved.

In this chapter, a different suboptimal excitation control design technique is presented. The linear optimal excitation control of SSR is designed in the usual way, but the controller is simplified by rejecting some feedback signals which have the least effect on the system damping, determined from an eigenvalue sensitivity analysis. Formulation will be given, suboptimal excitation control will be designed, and the results of both eigenvalue analysis and time domain simulation of the SSR control of a power system will be presented.

#### 4.2 Linear Optimal Control

For the linear optimal excitation control design, let the state equation be.

$$[x] = [A][x] + [B][u]$$
 (4.1)

and a cost index be chosen as

 $J = \frac{1}{2} \int_{0}^{\infty} ( [x]^{T} [Q] [x] + [u]^{T} [R] [u] ) dt \qquad (4.2)$ where [x] is the state variable vector, [u] the control vector, [A] the system matrix, [B] a control matrix, [Q] a positive semidefinite weighting matrix, and [R] a positive definite weighting matrix.

A Hamiltonian is formed by appending (4.1) to (4.2)

$$H = \frac{1}{2} ([x]^{T}[Q][x] + [u]^{T}[R][u]) + [p]^{T} ([A][x] + [B][u])$$
(4.3)

where [p] is the costate vector or Lagrange multipliers, and the optimal control can be found from  $\partial H/\partial u$ , resulting in

$$[u] = -[R]^{-1}[B]^{T}[p]$$
(4.4)

Let 
$$[p] = [K][x]$$
 (4.5)

and assume a time-invariant system, [ K ] must satify the following matrix Riccati equation

$$[K] [A] + [A]^{T} [K] - [K] [B] [R]^{-1} [B]^{T} [K] + [Q] = 0$$
 (4.6)

From the state equation of [x] and the co-state equation of [p], a composite system matrix [ M ] becomes

$$M ] = \begin{pmatrix} [A] & -[B] [R]^{-1} [B]^{T} \\ & & \\ -[Q] & -[A]^{T} \end{pmatrix}$$
(4.7)

There are 2n eigenvalues of matrix [ M ] for an n-th order system, and the eigenvalues are symmetrically distributed on the right and the left parts of the complex plane. Let the eigenvalue matrix be

$$[\Lambda] = \begin{bmatrix} \Lambda_{I} \\ & & \\ & & & \\ & & & & \\ & & & & & \\ & & & & & & \\ & & & & & & \\ & & & & & & \\ & & & & & & \\ & & & & & & & \\ & & & & & & & \\ & & & & & & & \\ & & & & & & & \\ & & & & & & & \\ & & & & &$$

and the corresponding eigenvector matrix be

$$[X] = \begin{pmatrix} X_{I} & X_{III} \\ & & \\ X_{II} & X_{IV} \end{pmatrix}$$
(4.9)

The Riccati matrix [K] may be computed from

$$[K] = [X_{II}] [X_{I}]^{-1}$$
(4.10)

where [ $\Lambda_{I}$ ] constitutes the n eigenvalues of [M] on the left hand side of the complex plane, which are the eigenvalues of the closed loop system matrix [ $\Lambda_{c}$ ], as the closed loop state equations may be written

$$[x] = [A][x] + [B][u]$$
  
= ([A] - [B][R]<sup>-1</sup>[B]<sup>T</sup>[K])[x] (4.11)  
= [A][x]

## 4.3 Eigenvalue Sensitivity [ 31,32 ]

Consider the controlled system matrix  $A_{c}$ . For the i-th eigenvalue  $\lambda_{i}$  and eigenvector  $X_{i}$ , we have

$$A_{c} X_{i} = X_{i} \lambda_{i}$$
(4.12)

For the i-th eigenvalue  $\lambda_i$  and the eigenvector  $V_i$  of the transposed  $A_c$ , we have

$$A_{c}^{T} V_{i} = V_{i} \lambda_{i}$$
(4.13)

Taking the partial derivative of both sides of equation (4.12) with respect to a system parameter  $\alpha$  gives

$$\frac{\partial A_{c}}{\partial \alpha} X_{i} + A_{c} \left( \frac{\partial X_{i}}{\partial \alpha} \right) = \lambda_{i} \left( \frac{\partial X_{i}}{\partial \alpha} \right) + \left( \frac{\partial \lambda_{i}}{\partial \alpha} \right) X_{i}$$
(4.14)

Premultiplying both sides of equation (4.14) by  $V_j^T$  results in

$$v_{j}^{T} \left( \frac{\partial A_{c}}{\partial \alpha} \right) x_{i} + v_{j}^{T} A_{c} \left( \frac{\partial X_{i}}{\partial \alpha} \right) = v_{j}^{T} \lambda_{i} \left( \frac{\partial X_{i}}{\partial \alpha} \right) + \left( \frac{\partial \lambda_{i}}{\partial \alpha} \right) v_{j}^{T} x_{i}$$
 (4.15)

But

$$V_{j}^{T} A_{c} = V_{j}^{T} \lambda_{j}$$
(4.16)

Therefore equation (4.15) becomes

$$U_{i}^{T} \left( \frac{\partial A_{c}}{\partial \alpha} \right) X_{i} = \left( \frac{\partial \lambda_{i}}{\partial \alpha} \right) V_{i}^{T} X_{i}$$
 (4.17)

or

$$\frac{\partial \lambda_{i}}{\partial \alpha} = \frac{v_{i}^{T} \left(\frac{\partial A_{c}}{\partial \alpha}\right) x_{i}}{\left(x_{i}, v_{i}\right)}$$
(4.18)

where ( $X_i$ ,  $V_i$ ) equals to  $V_i^T$ .  $X_i$ , according to (4.12) and (4.16)

## 4.4 Reduced Order Controller via Eigenvalue Sensitivity Analysis [24]

Let the feedback matrix  $[BR^{-1}B^{T}K]$  of (4.11) be simply written as [F], and let  $\Delta V_{R}$  fo the voltage regulator be the last state variable. The closed loop system matrix becomes

$$\begin{bmatrix} A_{c} \end{bmatrix} = \begin{bmatrix} A - F \end{bmatrix}$$
(4.20)  
$$\begin{bmatrix} F \end{bmatrix} = \begin{bmatrix} 0 \end{bmatrix}$$
(4.21)  
$$\begin{bmatrix} f_{n1}, \dots, f_{nk}, \dots, f_{nn} \end{bmatrix}$$

where

Therefore, the eigenvalue shift of the controlled system is due to the change in the last row of [F]. Since not all the feedback elements,  $f_{ni}$ , i=1,2,..,k,..,n, contribute substantial damping to the system, those which have relatively small contribution may be neglected, which will not affect the overall performance of the controlled system.

Since

therefore (4.18) becomes

$$\frac{\partial \lambda_{i}}{\partial f_{nk}} = \frac{-x_{i}^{k} \cdot v_{i}^{n}}{(x_{i}, v_{i})}$$
(4.23)

where  $X_i^k$  is the k-th element of the eigenvector  $X_i$ , and  $V_i^n$  is the n-th element of the eigenvector  $V_i$ .

(4.19)

(4.22)

When deleting a feedback element  $f_{nk}$ ,  $\Delta f_{nk} = -f_{nk}$ , therefore

$$\Delta \lambda_{i} \simeq \frac{X_{i}^{k} \cdot V_{i}^{n}}{(X_{i}, V_{i})} f_{nk} \qquad (4.24)$$

By examining  $\Delta\lambda_i$ , i=1,2,...,n, one can decide which feedback elements of [F] in (4.21) can be deleted.

The total effect of deleting some feedback elements can be cal-

$$\Delta \lambda_{i_{\text{total}}} \simeq \sum_{d} \frac{X_{i}^{k} \cdot V_{i}^{n}}{(X_{i}, V_{i})} f_{nd} \qquad (4.25)$$

where f 's are those feedback control elements being eliminated. Of course, one has to check whether

Re 
$$(\lambda_i + \Delta \lambda_i) < 0$$
 i=1,2,...,n (4.26)  
itotal

in order to have a stable system, where  $\lambda_i$ , i=1,2,...,n, are the eigenvalue of the controlled system without deleting any feedback signals.

## 4.5 Examples of the Controller Design

Two examples of the SSR control design are given in this section using the reduced order one-machine infinite-bus models developed in Chapter 3. For the 14th order model, the state variables are

$$[x] = [\Delta \omega_{IP}, \Delta \theta_{IP}, \Delta \omega_{LPB}, \Delta \theta_{LPB}, \Delta \omega, \Delta \delta, \Delta I_d, \Delta I_q, \Delta I_f, \Delta I_kd, \Delta I_{kq}, \Delta V_{cd}, \Delta V_{cq}, \Delta V_R]^T$$
(4.27)

For the excitation control design, [ B ] become a vector with only one non-zero element  $K_{\rm A}^{}/T_{\rm A}^{}$  associated with  $\Delta V_{\rm R}^{}$  .

From earlier experience [ 18 ], the weighting matrices are chosen as

$$[Q] = diag[5000,50,55000,50,3000,25;1000,1000, 0,5,5,1,1,0]$$
 (4.28)

[R] = 1

After design, the control is substituted into the original 26th order full model whose state variables are

$$[\mathbf{x}] = [\Delta\omega_1, \Delta\theta_1, \Delta\omega_2, \Delta\theta_2, \Delta\omega_3, \Delta\theta_3, \Delta\omega_4, \Delta\theta_4, \Delta\omega_5, \Delta\theta_6, \Delta\theta_6, \Delta\mathbf{a}, \Delta\mathbf{P}_{CV},$$

which corresponds to equations (2.41) and (2.42). Eigenvalue sensitivity technique is then applied. Typical eigenvalue shift due to individual state feedback is shown in Table 4.1.

#### Table 4.1

Typical value of the eigenvalue shift due to individual state feedback

Mechanical modes of the full order controlled system	Net eigenvalue shift due to the deletion of f25,9	
-0.1836 ± j298.18	0.0	
-0.6618 ± j203.32	+0.1182 ∓ j0.0641	
-0.1949 ± j160.68	-0.1228 ∓ j0.1891	
-0.6980 ± j127.08	-0.0033 ∓ j0.0334	
-0.2649 ± j98.889	-0.1053 ∓ j0.2758	
-3.1529 ± j4.8785	-0.5931 ∓ j0.2181	

It is found that the state feedback of  $\Delta \omega_{\mathrm{IP}}$ ,  $\Delta \theta_{\mathrm{IP}}$ ,  $\Delta \theta_{\mathrm{LPA}}$ ,  $\Delta \omega$ ,  $\Delta I_{\mathrm{kq}}$ ,  $\Delta V_{\mathrm{cd}}$ ,  $\Delta V_{\mathrm{cq}}$  do not have significant effect on the eigenvalues and hence may be deleted, resulting in a 7th order controller as follows

$$U_{\rm E} = 918.63\Delta\omega_{\rm LPB} - 46.159\Delta\delta + 214.58\Delta I_{\rm d} - 23.06\Delta I_{\rm q}$$
  
-200.67 $\Delta I_{\rm f}$  - 198.67 $\Delta I_{\rm kd}$  - 0.0558 $\Delta V_{\rm R}$  (4.30)

With similiar procedures, an excitation control is also designed using the 16th order model. The state variables of the model

$$\mathbf{x} ] = [\Delta \omega_{\mathrm{IP}}, \Delta \theta_{\mathrm{IP}}, \Delta \omega_{\mathrm{LPA}}, \Delta \theta_{\mathrm{LPA}}, \Delta \omega_{\mathrm{LPB}}, \Delta \theta_{\mathrm{LPB}}, \Delta \omega, \Delta \delta; \\ \Delta \mathbf{I}_{\mathrm{d}}, \Delta \mathbf{I}_{\mathrm{q}}, \Delta \mathbf{I}_{\mathrm{f}}, \Delta \mathbf{I}_{\mathrm{kd}}, \Delta \mathbf{I}_{\mathrm{kq}}, \Delta \mathbf{V}_{\mathrm{cd}}, \Delta \mathbf{V}_{\mathrm{cq}}, \Delta \mathbf{V}_{\mathrm{R}} ],$$

$$(4.31)$$

the weighting matrices are chosen as

$$[Q] = diag[500,0,10,1,600000,10,500,10;800,500, 0,0.5,0.5,1,1,0]$$
 (4.32)  
 $[R] = 1$ 

and the control  ${\rm U}_{\mathop{\rm E}}$  is found to be

$$U_{\rm E} = 54.896\Delta\theta_{\rm LPA} + 788.26\Delta\omega_{\rm LPB} - 11.187\Delta\theta_{\rm LPB} + 188.79\Delta I_{\rm d}$$
  
-9.725 $\Delta I_{\rm q}$  - 175.93 $\Delta I_{\rm f}$  + 8.4749 $\Delta I_{\rm kd}$  - 0.0404 $\Delta V_{\rm R}$  (4.33)

The operating conditions for which the above controllers

are designed are

electrical power = 0.9 per unit power factor = 0.9 lagging terminal voltage = 1.0 per unit capacitor compensation = 60%

## 4.6 The Output Feedback Control

Although the state variables  $\Delta \omega_{LPB}$ ,  $\Delta \delta$ , and  $\Delta \omega_{LPA}$  of the mass-spring system can be processed by the torsional stress analyzer [ 33 ], electrical variables  $\Delta I_d$ ,  $\Delta I_q$ ,  $\Delta I_{kd}^-$  can not be measured, and must be expressed in terms of measurable variables.

It is found that  $\Delta \Psi_{kq}$  is relatively small

 ${}^{\Delta\Psi}\mathbf{kq}$ 

(4.34)

$$\Delta \mathbf{I}_{\mathbf{kq}} \simeq \frac{\mathbf{X}_{\mathbf{mq}}}{\mathbf{X}_{\mathbf{kq}}} \Delta \mathbf{I}_{\mathbf{q}}$$

Since  $X_{mq}/X_{kq}$  is close to 1 for this particular case, the dynamic responses of  $\Delta I_q$  and  $\Delta I_{kq}$  as shown in Figure 4.1 are almost identical



Other equations are

$$P_{e} \approx \Psi_{d}I_{q} - \Psi_{q}I_{d}$$

$$Q_{e} \approx \Psi_{d}I_{d} - \Psi_{q}I_{q}$$

$$I_{t}^{2} = I_{d}^{2} + I_{q}^{2}$$
(4.35)

which are linearized for the SSR studies. Therefore  $\Delta I_d$ ,  $\Delta I_q$ ,  $\Delta I_{kd}$ ,  $\Delta I_f$  can be expressed in terms of the output variables  $\Delta P_e$ ,  $\Delta Q_e$ ,  $\Delta I_t$ ,  $\Delta I_f$  according to the following relation

(4.36)

Applying (4.36), the simplified controller of the 14th order model design of (4.30) becomes

$$U_{\rm E} = 918.63\Delta\omega_{\rm LPB} - 46.159\Delta\delta - 145.46\Delta P_{\rm e} - 68.665\Delta Q_{\rm e} + 160.88\Delta I_{\rm t} - 1.9982\Delta I_{\rm f} - 0.0558\Delta V_{\rm R}$$
(4.37)

which has seven measurable feedback signals, and that of the 16th order model design of (4.33) becomes

$$U_{\rm E} = 54.896\Delta\theta_{\rm LPA} + 788.28\Delta\omega_{\rm LPB} - 11.187\Delta\theta_{\rm LPB} - 118.85\Delta P_{\rm e}$$
$$-64.489\Delta Q_{\rm e} + 137.89\Delta I_{\rm t} - 1.0512\Delta I_{\rm f} - 0.0404\Delta V_{\rm R}$$
(4.38)

which has eight measurable feedback signals.

#### 4.7 Eigenvalue Analysis of the Simplified Controllers

The excitation controls of the 16th order model and 14th order model design are substituted into the linearized 26th order full model for eigenvalue analysis, for a wide range compensation of  $X_c/X_L$ from 10% to 80%, and for three operating conditions of  $P_e$  equal to 0.5, 0.9, and 1.25 per unit. The system with either one of these two controls under all conditions are stable. Typical results are given in Tables 4.2, 4.3, and 4.4. Note that the control designed for the 14th order model not only stabilizes mode 3 and mode 4, but also provides damping to mode 1 which is not included in the model for the design.

#### 4.8 Dynamic Performance Test using Nonlinear Model

The simplified controllers are also substituted into the nonlinear 26th order full model derived in Chapter 2 for dynamic performance test. A pulsed torque disturbance of 20% for 0.2 second is assumed for the system at various compensation and operating conditions. Typical responses of the system without and with control of the 14th order model design, at 0.9 per unit generator load and 50% capacitor compensation, are shown in Figures 4.2 and 4.3 respectively.

Although some responses of the system without control are unstable, responses of the system with control are all stable. Note that all responses are in per unit except the torque angle which is in degree. The speed response is the deviation from its steady state value.

# Table 4.2

Typical mechanical modes of the system with and without control at normal load (  $P_e = 0.9 \text{ p.u.}, Q_e = 0.4359 \text{ p.u.}, V_t = 1.0 \text{ p.u.}$ )

			المستجربين والمستجر والمستجر والمستجر والمشتقي والمستجر والمستجر والمستجر والمستجر والمستجر والمستجر
Compensation	Without the excitation control	With a control designed for the 16th order	With a control designed for the 14th order
	-0.1818±j298.18	-0.1819±j298.18	-0.1818±j298.18
	-0.4938±j203.60	-1.2264±j201.18	-0.8842±j201.52
	-0.2513±j160.64	-0.3396±j160.39	-0.2672±j160.45
30%	-0.6705±j127.02	-0.7310±j127.15	-0.6927±j127.09
	-0.2811±j99.136	-0.7864±j99.290	-0.4608±j99.295
	-0.0031±j8.4105	-1.3534±j5.1525	-1.3517±j6.3035
	-0.1818±j298.18	-0.1819±j298.18	-0.1818±j298.18
	+0.1237±j202.87	-0.4237±j203.56	-0.4554±j203.44
	+0.2603±j161.38	-0.6495±j160.11	-0.4192±j160.36
50%	-0.6828±j127.05	-0.7958±j127.23	-0.7322±j127.16
	-0.3528±j99.345	-1.0005±j99.479	-0.5894±j99.490
	-0.2327±j9.4692	-1.7884±j6.0755	-1.6931±j7.1144
	-0.1818±j298.18	-0.1819±j298.18	-0.1818±j298.18
	+0.0612±j202.88	-0.4547±j203.32	-0.4526±j203.22
	+0.0908±j160.45	-0.5202±j161.06	-0.3562±j160.72
60%	-0.6947±j127.23	-0.8516±j127.37	-0.7674±j127.25
	-0.4162±j99.567	-1.1792±j99.678	-0.6892±j99.674
	-0.3773±j10.099	-2.0708±j6.5702	-1.9159±j7.5326
	-0.1818±j298.18	-0.1819±j298.19	-0.1818±j298.18
	-0.0134±j202.89	-0.5034±j203.17	-0.4788±j203.09
	-0.0967±j160.52	-0.3236±j160.68	-0.2693±j160.58
80%	-0.5998±j126.94	-0.2165±j127.36	-0.2864±j127.16
	+1.7178±j102.17	-1.1561±j101.12	-0.6186±j100.85
	-0.7619±j11.662	-2.8558±j7.5796	-2.5196±j8.2862
· ·			

## Table 4.3

Typical mechanical modes of the system with and without control at light load (  $P_e = 0.5 \text{ p.u.}, Q_e = 0.4359 \text{ p.u.}, V_t = 1.0 \text{ p.u.}$ )

7				
-	Compensation	Without the Excitation control	With a control designed for the l6th order	With a control designed for the l4th order
	· .	-0.1818±j298.18	-0.1819±j298.18	-0.1819±j298.18
		-0.1624±j203.39	-0.6431±j203.39	-0.4097±j201.66
		-0.2079±j160.64	-0.2565±j160.45	-0.2110±j160.50
	30%	-0.6629±j127.02	-0.7198±j127.11	-0.6883±j127.07
		-0.2178±j99.147	-0.6132±j99.198	-0.3883±j99.227
		-0.4313±j8.4705	-1.4887±j5.8854	-1.3734±j6.4130
		-0.1818±j298.18	-0.1819±j298.18	-0.1819±j298.18
		+0.0503±j202.88	-0.5279±j203.59	-0.5680±j203.47
	50%	+0.3037±j160.06	-0.3573±j160.11	-0.1995±j160.34
		-0.6665±j127.05	-0.7922±j127.16	-0.7302±j127.12
		-0.2474±j99.356	-0.7957±j99.309	-0.4974±j99.381
		-0.6297±j9.4819	-1.9004±j6.6909	-1.7489±j7.1128
		-0.1818±j298.18	-0.1819±j298.18	-0.1819±j298.18
	:	+0.0138±j202.88	-0.4872±j203.39	-0.4972±j203.38
	60%	-0,0050±j160,42	-0.7074±j160.90	-0.4829±j160.56
		-0.6586±j127.11	-0.8700±j127.28	-0.7748±j127.20
		-0.2655±j99.553	-0.9639±j99.425	-0.5871±j99.520
		-0.7548±j10.082	-2.1637±j7.1201	-1.9926±j9.4656
		-0.1818±j298.18	-0.1818±j298.18	-0.1819±j298.18
		-0.0135±j202.91	-0.4812±j203.26	,-0.4728±j203.17
80%	80%	-0.1284±j160.52	-0.3631±j160.61	-0.3139±j160.59
		-0.6624±j126.94	-0.2603±j127.42	-0.3162±j127.19
		+1.7579±j101.46	-1.3391±j100.52	-0.6494±j100.47
		-1.0877±j11.568	-2.9051±j7.9656	-2.6895±j8.0386
_				1

Table 4.4

Typical mechanical modes of the system with and without control at heavy load (  $P_e = 1.25 \text{ p.u.}, Q_e = 0.4359 \text{ p.u.}, V_t = 1.0 \text{ p.u.}$ )

_				
	Compensation	Without the excitation control	With a control designed for the 16th order	With a control designed for the 14th order
		-0.1818±j298.18	-0.1819±j298.18	-0.1819±j298.18
		-0.8632±j203.72	-1.5884±j201.03	-1.1821±j201.42
	30%	-0.2907±j160.63	-0.3915±j160.33	-0.3036±j160.42
		-0.6774±j127.02	-0.7369±j127.17	-0.6952±j127.11
		-0.3384±j99.104	-0.8755±j99.329	-0.4942±j99.318
+		+0.4213±j8.1027	-1.4984±j4.7420	-1.5920±j6.4917
		-0.1818±j298.18	-0.1819±j298.18	-0.1819±j298.19
		+0.1881±j202.89	-0.3394±j203.55	-0.3682±j203.44
1	50%	+0.2020±j161.68	-0.8504±j160.10	-0.5762±j160.37
		-0.6981±j127.06	-0.7978±j127.26	-0.7343±j127.17
		-0.4477±j99.318	-1.1084±j99.570	-0.6348±j99.541
		+0.1360±j9.2235	-1.8497±j5.9294	-1.8564±j7.5255
		-0.1818±j298.18	-0.1819±j298.18	-0.1819±j298.18
		+0.1031±j202.89	-0.4175±j203.29	-0.4098±j203.19
	60%	+0.1726±j160.49	-0.4076±j161.16	-0.2738±j160.83
		-0.7293±j127.13	-0.8409±j127.41	-0.7656±j127.28
		-0.5537±j99.546	-1.2920±j99.823	-0.7407±j99.753
		-0.0402±j9.8917	-2.1070±j6.5627	-2.0503±j8.0835
		-0.1818±j298.18	-0.1819±j298.18	-0.1819±j298.18
		+0.0373±j202.91	-0.5021±j203.12	-0.4692±j203.05
	80%	-0.0687±j160.52	-0.2926±j160.67	-0.2361±j160.58
		-0.5800±j126.95	-0.1939±j127.33	-0.2693±j127.14
		+1.6687±j102.81	-0.9919±j101.47	-0.5752±j101.08
		-0.4977±j11.552	-2.8669±j7.9327	-2.6189±j9.2738
				~





Figure 4.2 (continued )


Figure 4.2 ( continued )



# Figure 4.2 (continued)





Figure 4.3 ( continued )



Figure 4.3 ( continued )



Figure 4.3 ( continued )

### 4.9 The Control Signal

The control signal  $U_E$  of the 14th order model design and its frequecy spectrum from Fourier analysis are shown in Figures 4.4 and 4.5 respectively. There are four significant peaks at 1.15 Hz , 15.7 Hz , 25.6 Hz , and 32.3 Hz in the spectrum, which correspond to mode 0, mode 1, mode 3, and mode 4 of the generator mass-spring system respectively. The stabilizing effects on all mode of oscillations of this design are well coordinated.

Although the output feedback excitation control of SSR proves very effective for a single machine system, its effectiveness on multimachine system is still unknown. The investigation of SSR control of multi-machine system with excitation control will be continued in the subsequent chapter.



### 5 MULTI-MACHINE SSR STUDIES

### 5.1 Introduction

The SSR literatures so far are dealing with the torsional oscillations and counter measures of single generating unit possibly for two reasons: one, it has been considered as a local problem, two, it is difficult to deal with a very high order multi-machine system.

In this chapter, the multi-machine SSR problem will be examined, and excitation control of SSR will be developed. Two working examples will be given: a two-machine system and a three-machine system. Two factors make SSR studies in a multi-machine system different from that of a single machine system. First, more than one electrical resonance frequency may exist for the series-capacitor-compensated multiple transmission lines. Second, the torsional interaction of the mass-spring systems and the dynamic interaction of the low frequency oscillations between machines may exist. Therefore, the strategy of controller design depends very much upon the degree of interaction between machines.

The procedures of the multi-machine SSR studies are as follows: first, the system is given an eigenvalue analysis to find the effect of other capacitor-compensated lines not directly connected to a particular machine on the torsional modes of that machine. Second, the torsional interacting effects of other machines on the individual systems are examined. Third, excitation controls of SSR are designed, using the techniques developed in the previous chapters, for the machines with unstable torsional modes. Finally, the system with controller is evaluated using eigenvalue analysis and computer simulation test using the nonlinear power system model.

A two-machine system for the SSR studies is shown in Figure 5.1 with components listed in Table 5.1 and data given in Appendix II. As for the operating conditions, the conditions of machine 1 are fixed ( i.e.  $P_{e1} = 0.9 \text{ p.u.}$ , P.F. = 0.9 lagging,  $V_{t1} = 1.0 \text{ p.u.}$ ) and these of machine 2 vary with the compensation level ( 10% - 80% ). The operating conditions of all buses other than the terminal bus of machine 1 are calculated for various compensation levels.



Figure 5.1 A two-machine power system.

	sys	No. of state	
	six mass-	spring system	12
Machine 1	five-wind:	ing generator model	5
	second ord	2	
	10		
Machine 2	five-wind:	5	
· ·	second ord	2	
	Line No.	capacitor compensated	
Transmission	1	yes	
system	2	yes	6
	3 and 4	no	
	<b></b>	Total No. of state	42

Summary of the components and number of state in the twomachine system

Table 5.1

A three-machine system for SSR studies is shown in Figure 5.2 with components listed in Table 5.2 and data given in Apprendix III. Three base cases are studied:

1) 50% compensation for all lines, except no compensation for line 7.

2) 60% compensation for all lines, except no compensation for line 7.

3) The compensation of line 1 and 5 is 40%, that of lines 3,4,6,and 8

is 70%, and that of line 9 is 35%. No compensation for line 7.

In addition to the three base case studies, line compensation is also varied one at a time in the range of 30% to 70% for further studies. Operating conditions of the three base case studies are given in Table 5.3.





### Table 5.2

Summary of the components and number of state in the three-machine system

	syst	No. of state	
	six mass-	spring system	12
Machine l	five-wind:	ing generator model	5
	second ord	ler excitation system	2
	five mass	10	
Machine 2	five-wind	ing generator model	5
	second or	2	
	four mass-spring system		8
Machine 3	five-winding generator model		5
~	second order excitation system		2
	Line No.	capacitor compensated	
Transmission	1 to 6	yes	28
system	7	no	
	8 to 9	yes	·
		Total No. of state	79

# Table 5.3

Various machine operating conditions in the three-machine system

		Machine 1	Machine 2	Machine 3
	Pe	0.9	0.9068	0.5
base case l	Q <sub>e</sub>	0.3359	0.3408	0.2744
	v <sub>t.</sub>	1.0 <u>-3.12</u> °	1.0 <u>/0°</u>	1.0 <u>-13.6</u> °
	Р е	0.9	0.9056	0.5
base case 2	Q <sub>e</sub>	0.3071	0.3196	0.2602
	v <sub>t.</sub>	1.0 /-2.38°	$1.0/0^{\circ}$	1.0 <u>/-11.48°</u>
	Р <sub>е</sub>	0.9	0.9086	0.5
base case 3	Q <sub>e</sub>	0.3076	0.3799	0.2627
	v <sub>t</sub>	$1.0 \  -4.56^{\circ}$	1.0 <u>0</u> °	1.0 <u>/-11.32</u> °

### 5.3 Prelimary Study of the two-machine System

From full model eigenvalue analysis of the two-machine system over a wide range of capacitor compensation, five unstable modes of the system are identified and given in Table 5.4 . Typical mechanical modes of the system are shown in Table 5.5 . The effect of capacitor compensation on the torsional modes of the machine not directly connected to that transmission line is also investigated. Typical results as manifested by the variation of real parts of the eigenvalues of particular torsional modes are shown in Figures 5.3 through 5.6 . The natural frequencies also change slightly. Therefore, for a multi-machine system with multiple capacitor-compensated-lines, more than one condition at which SSR may occur.

Table 5.4

Unstable modes of the two-machine system

	mode 3 (25.5 Hz.)
rachine i	mode 4 (32.3 Hz.)
	mode 0 ( 1-2 Hz.)
Machine 2	mode 2 (24.0 Hz.)
	mode 3 (30.2 Hz.)

Interaction between machines' torsional modes is obviously an important factor to be considered in the SSR controller design. If the interaction between machines is significant it must be considered, otherwise, the controller can be designed one machine at a time.

The torsional interaction between machines are investigated as follows:

1) Find the eigenvalues from the full model of the system.

2) Find the eigenvalues of the reduced order models that the massspring system of one machine is represented in detail and the mass-spring system of the other is lumped into one mass.

3) Compare the results of steps 1) and 2). Any significant change in the torsional modes indicates the existence of interaction. Otherwise, the interaction between machines is insignificant.

Typical results for the two-machine system are shown in Table 5.6. Although there are slight changes in the torsional modes between steps 1) and 2) especially when two mode frequencies are close, these interactions are not strong enough to make the unstable modes stable or vice versa. Therefore, the interaction. between the two machines' torsional modes is insignificant.

## Table 5.5

Typical mechanical modes of the two-machine system

Line compensation	Machine 1	Machine 2	
Line 1 70% Line 2 70%	-1.2096±j298.18 +0.1503±j203.02 -0.2213±j160.46 -0.7152±j127.08 -0.6302±j99.416 -0.9949±j10.242	-0.1215±j276.41 +0.0862±j189.96 +0.3546±j151.55 -0.2503±j102.42 -0.1677±j6.8353	
Line 1 70% Line 2 30%	-1.2224±j298.18 +0.1073±j202.87 +0.1379±j160.88 -0.7183±j127.08 -0.6550±j99.429 -0.6239±j10.422	-0.1215±j276.41 +0.1396±j189.81 -0.2186±j151.65 -0.2048±j102.26 +0.3318±j6.6557	
Line 1 50% Line 2 50%	-1.2066±j298.18 +0.0610±j202.91 -0.5470±j160.75 -0.7039±j99.204 -0.6625±j9.3108	-0.1215±j276.41 +0.1352±j189.97 -0.1927±j151.52 -0.0119±j6.5236	
Line 1 50% Line 2 70%	-1.1997 ±j298.18 -0.0383 ±j203.38 -0.4321 ±j160.95 -0.7027 ±j127.04 -0.5727 ±j99.211 -0.9671 ±j9.3253	-0.1215 ±j276.41 -0.0467 ±j189.98 -0.0619 ±j151.68 -0.2159 ±j102.27 -C.1467 ±j6.6456	

### Table 5.6

Typical mechanical modes of the two-machine system using different models for the investigation of torsional interaction between machines

		Ml : detail M2 : detail	Ml : detail M2 : one mass	Ml : one mass M2 : detail
Line 1 60%	Machine 1 (Ml)	-1.2143±j298.18 +0.1296±j202.87 -0.5487±j161.02 -0.7096±j127.05 -0.6088±j99.282 -0.6064±j9.7300	-1.2143±j298.18 +0.1303±j202.87 -0.5472±j160.01 -0.7095±j127.05 -0.6179±j99.299 -0.6064±j9.730	-0.5022±j9.7802
Line 2 40%	Machine 2 (M2)	-0.1215±j276.41 -0.0295±j189.95 -0.1843±j151.52 -0.1887±j102.18 -0.0831±j6.4809	-0.0799±j6.5063	-0.1215±j276.41 -0.0296±j189.95 -0.1850±j151.52 -0.1797±j102.16 -0.0623±j6.4929
Line l 70%	Machine l (M1)	-1.2194±j298.18 +0.0714±j202.87 +0.1326±j160.77 -0.7167±j127.08 -0.6448±j99.411 -0.6822±j10.288	-1.2914±j298.18 +0.0718±j202.87 +0.1228±j160.78 -0.7167±j127.07 -0.6612±j99.438 -0.6944±j10.313	-0.5805±j10.351
Line 2 40%	Machine 2 (M2)	-0.1215±j276.41 -0.0509±j189.95 -0.1115±j151.62 -0.1970±j102.25 -0.1094±j6.7660	-0.1065±j6.7949	-0.1215±j276.41 -0.0509±j189.96 -0.1058±j151.96 -0.1807±j102.23 -0.0885±j6.7772







Figure 5.4

Variation of the real part of mode 4 (203 rad/sec.) of machine 1 (vertical axis) as line 2's capacitor changes and line 1's capacitor kept constant at (a) 10%, (b) 20%, (c) 30%, (d) 40%, (e) 50%, (f) 60%, (g) 70%, (h) 80% compensation.

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Figure 5.5

Variation of the real part of mode 2 (151 rad/sec) of machine 2 (vertical axis) as line 1's compensation changes and line 2's capacitor kept constant at (a) 10%, (b) 20%, (c) 30%, (d) 40%, (e) 50%, (f) 60%, (g) 70%, (h) 80% compensation.



Figure 5.6

Variation of the real part of mode 3 (190 rad/sec.) of machine 2 (vertical axis) as line 1's capaciotr changes and line 2's capacitor kept constant at (a) 10%, (b) 20%, (c) 30%, (d) 40%, (e) 50%, (f) 60%, (g) 70%, (h) 80% compensation.

### 5.4 Preliminary Study of the Three-machine System

Again from the full model eigenvalue analysis of the threemachine system over a wide range of capacitor compensation, six unstable modes of the system are identified, and they are given in Table 5.7 . Typical mechanical modes of the system are given in Table 5.8 .

The interacting effect of torsional modes for the system is also investigated using the same procedures as described in section 5.3. The results are shown in Table 5.9, which also suggest that interaction between machine's torsional modes is insignificant.

Tabl	.e	5	.7
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Unstable modes of the three-machine system

Machine 1	mode 1 (15.7 Hz.)
	mode 4 (32.3 Hz.)
Machine 2	mode 0 ( 1-2 Hz.)
	mode 1 (16.2 Hz.)
	mode 2 (24.0 Hz.)
	mode 3 (30.2 Hz.)
Machine 3	none

# Table 5.8

Typical mechanical modes of the three-machine system

Line compensation	Machine 1	Machine 2	Machine 3
Line 1 60% Line 2 70% Line 3-9 60% (except 7)	-0.1818±j298.18 +0.2653±j203.29 -0.2753±j160.70 -0.6612±j127.04 -0.2455±j99.456 -1.7915±j10.363	-0.1214±j276.41 -0.0644±j189.99 +0.0832±j151.81 +0.2785±j102.65 +1.3101±j10.175	-0.1336 <sup>±</sup> j353.25 -0.1336 <sup>±</sup> j190.17 -0.7337 <sup>±</sup> j167.73 -3.2331 <sup>±</sup> j17.474
Line 1-8 50% (except 7) Line 9 30%	-0.1818±j298.18 -0.2359±j203.20 -0.2283±j160.66 -0.6773±j127.02 -0.2707±j99.224 -1.1930±j9.8449	-0.1215±j276.41 +0.1086±j189.99 -0.0544±j151.58 -0.3096±j102.66 +0.6526±j8.8284	-0.1344±j353.24 -0.2284±j190.11 -0.7334±j167.73 -3.1623±j17.262
Line 1-3 50% Line 4 30% Line 5-9 50% (except 7)	-0.1818±j298.18 +0.0608±j203.34 -0.3752±j160.63 -0.6427±j126.99 +0.1309±j99.502 -0.9892±j10.472	-0.1214±j276.41 -0.0549±j189.84 +0.1253±j151.93 -0.4555±j102.56 +0.2684±j8.7070	-0.1334±j353.24 -0.1776±j190.28 -0.7326±j167.73 -3.3444±j17.589
Line 1,5 40% Line 9 35% Line 2-4,6,8 70%	-0.1818±j298.18 +0.2653±j203.29 -0.2753±j160.70 -0.6612±j127.04 -0.2455±j99.456 -1.7915±j10.363	-0.1214±j276.41 -0.0644±j189.99 +0.0832±j151.81 +0.2785±j102.65 +1.3101±j10.175	-0.1336±j353.24 -0.1631±j190.17 -0.7337±j167.73 -3.2331±j17.474

Table	5	e	9
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Typical mechanical modes of the three-machine system using different models for the investigation of torsional interaction between machines

1				· · · · · · · · · · · · · · · · · · ·			
	Ml : detail	Ml : one mass	Ml : detail	Ml : detail	Ml : one mass	M1 : one mass	Ml : detail
	M2 : detail	M2 : detail	M2 : one mass	M2 : detail	M2 : one mass	M2 : detail	M2 : one mass
	M3 : detail	M3 : detail	M3 : detail	M3 : one mass	M3 : detail	M3 : one mass	M3 : one mass
	-0.182±j298.2		-0.182±j298.2	-0.182±j298.2			-0.182±j298.2
(TH	+0.265±j203.3		+0.253±j203.3	+0.266±j203.3			+0.253±j203.3
1	-0.275±j160.7		-0.270±j160.7	-0.275±j160.7			-0.270±j160.7
ne	-0.661±j127.0		-0.661±j127.0	-0.661±j127.0			-0.661±j127.0
ic hi	-0.245±j99.46		-0.412±j99.45	-0.245±j99.46			-0.411±j99.45
Ma	-1.792±j10.36	-1.818±j10.41	-1.805±j10.40	-1.792±j10.36	-1.834±j10.45	-1.818±j10.41	-1.805±j10.40
			-			·····	
[]	-0.121±j276.4	-0.121±j276.4		-0.121±j276.4	-	-0.121±j276.4	
<u>À</u>	-0.064±j189.9	-0.067±j189.9		-0.013±j190.0		-0.014±j190.0	
le ?	+0.083±j151.8	+0.080±j151.8	:	+0.083±j151.8		+0.081±j151.8	-
hin	+0.278±j102.6	+0.414±j102.6		+0.278±j102.6		+0.413±j102.6	
Mac	+1.310±j10.18	+1.327±j10.20	+1.331±j10.25	+1.310±j10.18	+1.350±j10.27	+1.327±j10.20	+1.332±j10.25
				· · · · · · · · · · · · · · · · · · ·			
(M3	-0.134±j353.2	-0.134±j353.2	-0.134±j353.2		-0.134±j353.2		
3	-0.163±j190.2	-0.161±j190.2	-0.216±j190.1		-0.216±j190.1		
ìn€	-0.734±j167.7	-0.734±j167.7	-0.734±j167.7		-0.734±j167.7	•	
lac!	-3.233±j17.47	-3.233±j17.47	-3.234±j17.48	-3.248±j17.51	-3.235±j17.47	-3.248±j17.51	-3.249±j17.51
2							•

#### 5.5 Controller Design Considerations of Multi-machine SSR System

Although the controller design using the full model, which accounts for all natural frequencies in the electrical system and torsional interaction between machines, could be the best, it often results in a high order controller. Since the torsional interaction between machines according to the foregoing studies is insignificant, an alternative method using a reduced order model for the design is developed. The original system is divided into several one-machine systems, and the controller is designed one at a time. Two general steps are as follows:

1) Choose an appropriate one-machine system model by retaining one of the transmission lines directly connected to the machine, which has the largest steady current and hence the strongest torsional interacting effect between the electrical and mechanical systems.

2) Modify the line by adding some impedance so that the critical electrical frequency for the line with compensation as viewed from the machine will not change.

When the multi-machine system is divided into several onemachine systems, the dynamic interaction between machines has been neglected. But these interactions depend very much upon the tie lines between machines. For a strong tie line , strong interaction may exist. Therefore, the controllers designed for individual one-machine systems must be coordinated. An iterative scheme as shown in Figure 5.7 may be applied to adapt the designed controllers so that all dampings of the mechanical modes in the system are coordinated.





Iterative scheme for adapting controller into the original system.

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### 5.6 Controllers Design and Test of the Two-Machine System

The two-machine system is divided into two one-machine infinite-bus system models for the controller design as follows:

1) The natural frequencies of either one of the two transmission lines for a wide range of capacitor compensation are determined by setting the capacitor compensation of the other line to zero.

2) The rest of the system is replaced by an equivalent impedance,  $X_{E1}$  or  $X_{E2}$  as shown in Figure 5.8(a) or (b) respectively, such that each system will have the same electrical natural frequencies as determined in step 1).



Figure 5.8 Two subsystems resulted from the two-machine system.

For machine 1 of Figure 5.8(a),  $X_{E1}$  is found to be 0.1 p.u. . Since the operating conditions, the machine and transmission line parameters of the system given in Appendix II are almost the same as that of the one-machine system studied in the previous chapters, the same controller in Equation (4.37) will be applied without change.

For machine 2 of Figure 5.8(b),  $X_{E2}$  is 0.11 p.u., and the average operating conditions are

 $P_{e2} = 0.8 \text{ p.u.}$ ,  $Q_{e2} = 0.4 \text{ p.u.}$ ,  $V_{t2} = 0.9 \text{ p.u.}$ 

Applying the mass-spring equivalencing technique developed in Chapter 3, the mass-spring system of machine 2 is reduced to a three-mass equivalent

retaining only modes 2 and 3. The mode shapes of the original and the equivalent system are shown in Figure 5.9. Including the electrical system, a reduced 14th order model is obtained. The state variables are

$$[x_{2}] = [\Delta \omega_{IP2}, \Delta \theta_{IP2}, \Delta \omega_{LPB2}, \Delta \theta_{LPB2}, \Delta \omega_{Gen2}, \Delta \delta_{Gen2}, \Delta I_{d2}, \Delta I_{q2}, \Delta I_{f2}, \Delta I_{kd2}, \Delta I_{kq2}, \Delta V_{cd2}, \Delta V_{cq2}, \Delta V_{R2}]$$
(5.1)

Using the 14th order model, an excitation control is designed using the linear optimal control laws and eigenvalue sensitivity technique developed in Chapter 4, resulting an 8th order controller.

The weighting matrices are

$$[R] = 1$$
  
 $[Q] = diag[5000,50,50000,50,100,25;100,100,$   
 $0,50,50,1,1,0]$  (5.2)

The state feedback control after the eigenvalue sensitivity analysis results in

$$U_{E2} = 633.82 \Delta \omega_{LPB2} + 35.559 \Delta \theta_{LPB2} - 56.804 \Delta \delta_{Gen2} + 56.63 \Delta I_{d2} + 12.453 \Delta I_{q2} - 56.089 \Delta I_{f2} - 52.01 \Delta I_{kd2} - 0.214 \Delta V_{R2}$$
 (5.3)

Applying (4.37), the state feedback control of (5.3), in terms of output variables, becomes

$$U_{E2} = 633.83 \Delta \omega_{LPB2} + 35.559 \Delta \theta_{LPB2} - 56.804 \Delta \delta_{Gen2} - 17.481 \Delta P_{e2}$$
  
-27.287 \Delta Q\_{e2} - 4.0792 \Delta I\_{f2} + 29.981 \Delta I\_{t2} - 0.214 \Delta V\_{R2} (5.4)



Figure 5.9 Mode shapes of the original and equivalent systems for machine 2 in the two-machine system.

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5.6.1 Testing of Controllers Using Eigenvalue Analysis

After the controller design, eigenvalues of the two-machine system are analyzed in the sequence as shown in Table 5.10.

#### Table 5.10

Testing sequence for the two-machine system

	Machine 1	Machine 2
1	with control	without control
2	without control	with control
3	with control	with control

The first two tests are used to examine the effect of the controller designed for one machine on the other machine, typical results are shown in columns 2 and 3 of Table 5.11 . It was found that the damping provided by the controller to the torsional modes of the other machine is insignificant. However, the effect of the controller on mode 0 of the other machine is noticable. It may give positive or negative damping to the other machine, depending on the operating conditions.

With both controllers applied to machines 1 and 2, all unstable modes are stabilized over the entire range of capacitor compensation. Typical results are shown in column 4 of Table 5.11 . Note that the controllers provide substantial amount of damping to mode 1 of both machines even though they are not considered in the controller design.

Although the effect of controller on mode 0 of the other machine is noticable, it is not significant due to the weak tie lines between the machines. Therefore, each control can be separately designed and the iterative scheme presented in Figure 5.7 is not necessary.

Table 5	.1	T
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Typical mechanical modes of the two-machine system without and with control

		Ml without	Ml with	Ml without	Ml with
		M2 without	M2 without	M2 with	M2 with
Line 1 50%, Line 2 30%	Machine 1 (M1)	-1.2128±j298.18 +0.4186±j202.99 -0.5580±j160.73 -0.7047±j127.03 -0.5844±j99.199 -0.4836±j9.2996	-1.2129±j298.18 -0.9341±j203.25 -0.5254±j160.51 -0.7329±j127.10 -0.7476±j99.349 -1.9182±j7.9942	-1.2128±j298.18 +0.3826±j203.00 -0.5551±j160.73 -0.7047±j127.03 -0.5813±j99.203 -0.5765±j9.4304	-1.2129±j298.18 -0.9107±j203.19 -0.5249±j160.51 -0.7328±j127.10 -0.7429±j99.351 -2.2889±j8.5922
	Machine 2 (M2)	-0.1214±j276.41 -0.0318±j190.02 -0.1690±j151.46 -0.1711±j102.11 -0.0261±j6.3218	-0.1214±j276.41 -0.0621±j190.02 -0.1665±j151.46 -0.1700±j102.11 -0.2053±j5.7251	-0.1223±j276.41 -0.4476±j189.17 -0.1992±j151.06 -0.6437±j102.06 -2.4631±j5.8497	-0.1223±j276.41 -0.3403±j189.15 -0.2045±j151.07 -0.6434±j102.23 -2.4449±j4.7145
, Line 2 70%	Machine 1 (M1)	-1.2096±j298.18 +0.1503±j203.02 -0.2213±j160.46 -0.7152±j127,08 -0.6302±j99.416 -0.9949±j10.246	-1.2096±j298.18 -0.4857±j203.03 -0.4882±j160.52 -0.7699±j127.19 -0.8320±j99.550 -3.0374±j9.0538	-1.2096±j298.18 +0.1422±j203.01 -0.2893±j160.53 -0.7116±j127.11 -0.6115±j99.439 -0.9926±j10.613	-1.2098±j298.18 -0.4790±j203.01 -0.5114±j160.52 -0.7673±j127.21 -0.8113±j99.565 -2.7396±j9.9798
Line 1 70%	Machine 2 (M2)	-0.1215±j276.41 +0.0862±j189.96 +0.3546±j151.51 -0.2503±j102.42 -0.1677±j6.8352	-0.1215±j276.41 +0.0779±j189.95 +0.2397±j151.80 -0.2336±j102.41 -0.0079±j6.0831	-0.1221±j276.41 -1.2334±j189.43 -0.8976±j151.13 -1.0867±j102.52 -2.3147±j6.8506	-0.1221±j276.41 -1.1587±j189.43 -1.0465±j150.77 -1.0463±j102.55 -2.2604±j5.4211

### 5.6.2 Dynamic Performance Test of the Two-machine System

The controller designs in (4.38) and (5.4) are substituted into the nonlinear two-machine system model for dynamic performance test. A three-phase fault for 0.075 second is assumed at the load bus as shown in Figure 5.10. Typical responses of the system with 50% compensation for both lines 1 and 2 without control are shown in Figures 5.11 and 5.12, and those with control in Figures 5.13 and 5.14 respectively.

For the system without control, most responses of the machines are either oscillatory or unstable. However, all responses of the system with excitation control are stable and all oscillations are damped out within 5 seconds.









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Figure 5.12 (continued)




Figure 5.13 (continued)





### 5.7 Controller Design and Test of the Three-machine System

Since all the mechanical modes of machine 3 are stable, the controller design will be focused on machines 1 and 2. Two one-machine models for machines 1 and 2, by retaining the strongest torsional interaction path and the critical electrical frequency for each model are chosen as follows:

1) Among the three transmission lines 1, 3, and 6 connected to machine 1, only line 6 is retained because it has the largest per unit current indicating the strongest interaction path. The rest of the system can be replaced by an equivalent reactance  $X_{F1}$  of 0.07 p.u..

2) For machine 2, only line 2 is retained and the rest of the system is replaced by an equivalent reactance  $X_{E2}$  of 0.1 p.u.

Again, the same controller of (4.37) is used for machine 1, because the operating conditions, the machine and transmission line parameters for machine 1 are almost the same as those of the system previously studied.

The operating conditions for machine 2 are

 $P_{e2} = 0.906 \text{ p.u.}, \quad Q_{e2} = 0.341 \text{ p.u.}, \quad V_{t2} = 1.0 \text{ p.u.}$ 

Using the 14th order model of (5.1) together with the techniques develpoed in Chapter 4, an 8th order excitation control for SSR is designed by choosing the same weighting matrices as shown in (5.2).

When both controllers are applied to the three-machine system, all unstable torsional modes in the system are stabilized over the entire range of the prescribed operating conditions. However, mode 0 of machine 2 remains unstable due to the inadequacy of the one-machine model by which the dynamic interaction between machines is neglected. But either one of the two controllers can be adapted also to stabilize mode 0 using the iterative scheme as shown in Figure 5.7.

## 5.7.1 Sensitivity Studies and Choice of Weighting Elements in [Q]

Dynamic interaction between machines is transmitted through the electrical network by the line current. Therefore, proper choice of weighting elements in conjunction with current is important to enhance the mode 0 damping in a multi-machine system.

For the three-machine system, the controller for machine 2, originally based on a one-machine model, is adapted by studying the sensitivity of mechanical damping with respect to the weighting elements of  $\Delta I_{kd2}$ ,  $\Delta I_{kq2}$ ,  $\Delta I_{d2}$ ,  $\Delta I_{q2}$ .  $\Delta I_{kd2}$  and  $\Delta I_{kq2}$  are included because they affect the self damping of machine 2 which in turn affect the dynamic interaction between machines. Keeping other weighting elements of [Q] in (5.2) constant, the effect of the weighting elements of  $\Delta I_{kd2}$  and  $\Delta I_{kq2}$  and  $\Delta I_{kq2}$  on the mechanical modes of machine 2 is shown in Figure 5.15 and that of  $\Delta I_{d2}$  and  $\Delta I_{a2}$  in Figure 5.16.

Two observations are as follows:

1) As the weighting elements  $Q_{Ikd2}$  and  $Q_{Ikq2}$  increase (more penalty on the deviation of damper winding currents are imposed ), mode 0 damping of machine 2 increases and that of machine 1 decreases. Damping of machine 2's torsional modes also decrease, but at a much slower rate.

2) As the weighting elements  $Q_{Id2}$  and  $Q_{Iq2}$  decrease (penalty on the deviation of stator currents is reduced ), mode 0 damping of machine 2 increases and that of machine 1 decreases. Damping of the torsional modes of machine 2 remains fairly constant.

Since the weighting elements of the damper winding currents affect the damping of machine 2's torsional modes, they must be chosen such that all mechanical modes in the three-machine system have reasonable positive damping. In this case  $Q_{Id2}$  and  $Q_{Iq2}$  of (5.2) are adapted to 50 ,  $Q_{Ikd2}$  and  $Q_{Ikq2}$  adapted to 100. The resulting weighting matrice are

$$[R] = 1$$
  
 $[Q] = diag[5000,50,50000,50,100,25;50,50,0$  (5.3)  
100,100,1,1,0]

and the designed state feedback control after simplification from eigenvalue sensitivity analysis becomes

$$U_{E2} = 549.26\Delta\omega_{LPB2} + 67.118\Delta\theta_{LPB2} - 80.172\Delta\delta_{Gen2} + 47.76\Delta I_{d2} + 12.289\Delta I_{q2} - 48.478\Delta I_{f2} - 41.527\Delta I_{kd2} - 0.249\Delta V_{R2}$$
(5.4)

Applying (4.36), the control of (5.4) in terms of output variables becomes

$$U_{E2} = 549.26\Delta\omega_{LPB2} + 67.118\Delta\theta_{LPB2} - 80.172\Delta\delta_{Gen2} - 14.041\Delta P_{e2}$$
  
-12.196\DeltaQ\_{e2} - 6.95\DeltaI\_{f2} + 99.296\DeltaI\_{t2} - 0.249\DeltaV\_{R2} (5.5)

With controller of (4.37) on machine 1 and (5.5) on machine 2 of the three-machine system, all unstable modes in the system are stabilized over the entire range of the prescribed operating conditions. Typical results are shown in Table 5.12.

# Table 5.12

Typical mechanical modes of the three-machine with control

Line compensatio	on	Machine 1	Machine 2	Machine 3
Line 1 60 Line 2 70 Line 3-9 60 (except 7)	0% 0% 0%	-0.1820±j298.18 -1.0612±j202.58 -0.3141±j160.49 -0.7058±j127.16 -0.5637±j99.549 -1.9871±j10.923	-0.1222±j276.41 -0.6101±j189.31 -0.3091±j151.04 -0.6588±j103.29 -1.3621±j5.4707	-0.1334±j353.24 -0.2016±j190.16 -0.7333±j167.73 -3.3618±j17.607
Line 1-8 50 (except 7) Line 9 30	0%	-0.1820±j298.18 -0.6224±j202.15 -0.2893±j160.48 -0.7071±j127.15 -0.5635±j99.507 -2.1187±j9.8608	-0.1221±j276.41 -0.4904±j189.45 -0.2752±j151.12 -0.9266±j102.71 -1.0805±j5.2511	-0.1337±j353.24 -0.2366±j190.09 -0.7334±j167.73 -3.1780±j17.277
Line 1-3 50 Line 4 30 Line 5-9 50 (except 7)	0% 0% 0%	0.1820±j298.18 -0.6799±j202.25 -0.3017±j160.31 -0.6629±j127.14 -0.5368±j99.606 -2.1820±j9.8276	-0.1221±j276.41 -0.3588±j189.41 -0.3372±j151.36 -0.9538±j102.42 -1.1479±j5.2261	-0.1335 <sup>±</sup> j353.24 -0.1876 <sup>±</sup> j190.05 -0.7313 <sup>±</sup> j167.73 -3.2170 <sup>±</sup> j17.439
Line 1,5 40 Line 9 35 Line 2-4,6,8 70	0% 5% 0%	-0.1820±j298.18 -1.5122±j202.49 -0.3139±j160.46 -0.6738±j127.17 -0.5616±j99.859 -1.9833±j11.218	-0.1221±j276.41 -0.5766±j189.39 -0.3147±j151.17 -0.2140±j102.85 -1.5807±j5.6033	-0.1336 <sup>±</sup> j353.24 -0.2310 <sup>±</sup> j190.14 -0.7336 <sup>±</sup> j167.73 -3.2409 <sup>±</sup> j17.489



# Figure 5.15

Variation of mechanical damping as weighting elements  $Q_{\rm Ikd2}$  and  $Q_{\rm Ikq2}$  change while  $Q_{\rm Id2}$  and  $Q_{\rm Iq2}$  constant at 50.





# Figure 5.16

Variation of mechanical damping as weighting elements  ${\rm Q}_{Id2}$  and  ${\rm Q}_{Iq2}$  change while  ${\rm Q}_{Ikd2}$  and  ${\rm Q}_{Ikq2}$  constant at 50.



Figure 5.16 (continued)

## 5.7.2 Dynamic Performance Test of the Three-machine System

The dynamic performance of the three-machine system without and with excitation control is tested using the nonlinear system model. A resistive load is switched into the system at bus 8 for 0.075 second as shown in Figure 5.17 , so that the bus voltage will drop 20% . Typical responses of the system without control are shown in Figures 5.18 through 5.20, and those with control in Figures 5.21 through 5.23 respectively. The line compensation of the system is 50% for all lines, except for line 7 and line 2 which has no compensation and 70% compensation respectively.

For the system without control, some responses of machines 1, 2, and 3 are unstable, but the responses of all machines are stable for the system with the excitation control.



Figure 5.17

The three-machine system subjected to disturbance.





Figure 5.18 (continued)

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Figure 5.19 (continued)





Figure 5.20 (continued)





Figure 5.21 (

(continued)





Figure 5.22 (continued)





Figure 5.23 (continued)

#### 5.8 Concluding Remarks for Multi-machine SSR Studies

From the foregoing SSR studies of the two-machine and threemachine systems, general conclusions are as follows:

- In a multi-machine system with multiple capacitor-compensated transmission lines, there is more than one condition at which SSR may occur.
- Interaction between torsional modes of different machines has no significant effect on SSR stability.
- 3) To apply the simplified output feedback excitation control design technique developed for the one-machine system to a multi-machine system, a one-machine equivalent including the strongest transmission tie and the critical electrical resonance frequency must be derived for each machine in the multi-machine system. When the controllers thus designed are applied to the multi-machine system, some adaptation may be required, which can be achieved using an iterative process.
- 4) Both eigenvalue analysis and dynamic performance test using nonlinear full models prove that the excitation control designed according to the procedures presented in this chapter is very effective for multimachine multi-mode stabilization of the SSR.

#### 6. CONCLUSION

Several useful model reduction, equivalencing, and control simplification techniques have been developed and many interesting results are found in this thesis study.

After presenting a unified electrical and mechanical model for the SSR studies in Chapter 2, it is shown in section 2.5 that, although the negative resistance concept is useful to explain the torsional interaction between the electrical and mechanical systems, the lumped mass representation of the turbine-generator is not sufficient, and the multimass-spring system must be used for SSR studies.

For the SSR control design, the excitable torsional modes are identified from modal analysis in Chapter 3. A mass-spring equivalencing technique is then developed for order reduction by retaining only the unstable torsional modes at certain frequencies, resulting in a lower order mass-spring system.

Based on the linear optimal control laws and with the reduced 14th and 16th order models of the original 26th order system, linear optimal excitation controllers are designed in Chapter 4. The controllers are further simplified from a sensitivity analysis by deleting some less sensitivity feedback signals, and the final controllers employ the system output signals as the feedback. Both eigenvalue analysis of the linearized full model and the computer simulated dynamic performance test using nonlinear full model indicate that the linear optimal excitation control thus designed is effective in providing damping to all torsional modes of the system over a wide range of capacitor compensation and operating conditions.

The stabilization technique is further extended and applied to a two-machine and a three-machine SSR systems in Chapter 5. It is found that the SSR stabilizers still can be designed for one machine at a time but coordination may be required after all controllers are implemented. For the individual machine SSR controller design, however, it is necessary to derive a one-machine equivalent for each individual machine by retaining the strongest tie of the machine, which has the largest current, to the remaining system, an equivalent reactance is adapted so that the electrical resonance frequency which affects the vulnerable torsional mode of the machine is retained. For the coordination of the damping provided by all controllers for the entire system, an iterative process is developed. Extensive eigenvalue analysis and nonlinear computer simulation tests again indicate that linear optimal excitation controls thus designed are very effective for the SSR control.

The linear optimal excitation developed in the thesis probably provides the most effective and least expensive means to stabilize the SSR of one-machine as well as multi-machine systems.

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#### APPENDIX I

#### SYSTEM DATA FOR THE ONE MACHINE SYSTEM

All the data are in per unit based on 500KV and 900 MVA except the time constant which is in second.

### Synchronous Machine Parameters

Xa	= .	1.79	x,	=	1.6999	R <sub>f</sub>	=	0.00105
X md	=-	1.66	X. kd	=	1.6657	R <sub>kd</sub>		0.00371
Xq	=	1.71	X kq	=	1.6531	R <sub>kq</sub>	=	0.00491
Xmq	=	1.58	Ra	=	0.0015			

Mass-spring System

M <sub>HP</sub>	=	0.185794			
 M	=	0.311178	<sup>к</sup> 12	=	19.303
IP			к <sub>23</sub>	=	34.929
<sup>M</sup> LPA	=	1.717340	K	tt	52.038
M <sup>1</sup> LPB	=	1.768430	34		70.050
M_	=	1.736990	<sup>K</sup> 45	F	70.858
Gen		0.0000	к <sub>56</sub>	=	2.8220
<sup>M</sup> Ex	=	0.068433			

# Turbine and Governing System

-									
	Кg	=	25	<sup>T</sup> 1	=	0.2	т <sub>2</sub>	=	0.0
	<sup>т</sup> з	=	0.3	т <sub>СН</sub>	=	0.3	т <sub>RH</sub>	<b>m</b>	7.0
	тсо	=	0.2	F <sub>HP</sub>	=	0.3	FIP	=	0.26
	F <sub>LPA</sub>	-=	0.22	F LPB	=	0.22	Р <sub>GV</sub>	=	±0.1

Exciter and Voltage Regulator

 $K_{A} = 50$  $T_{A} = 0.01$ 0.002 T<sub>E</sub> Exciter voltage ceiling limits =  $\pm 7.0$ 

## Transmission line Parameters

X <sub>t</sub>	=	0.14	R <sub>t</sub>	= 0.01	· ·	$\mathbf{x}_{\mathbf{L}}$	=	0.56
<sup>R</sup> L	-==	0.02	X <sub>c</sub> v	varied from	.0.056 -	0.448(3	10%	- 80%)

0.1

i=1,2,...,6

D<sub>ii</sub> =

#### APPENDIX II

SYSTEM DATA FOR THE TWO MACHINE SYSTEM

### A2.1 Machine 1

# Synchronous Machine Parameters

Similiar to that of the machine in Appendix I

## Mass-spring System

Similiar to that of the machine in Appendix I

#### Turbine Torque Distribution

$$F_{HP1} = 0.3$$
  $F_{IP1} = 0.26$   $F_{LPA1} = 0.22$   
 $F_{LPB1} = 0.22$ 

Exciter and Voltage Regulator

 $K_{A1} = 50$   $T_{A1} = 0.01$   $T_{E1} = 0.002$ 

Exciter voltage ceiling limits =  $\pm 7.0$ 

#### A2.2 Machine 2

Synchronous Machine Parameters

$X_{d2} = 1$	.82	x_=	1.92	<sup>R</sup> f2	=	0.0067
$X_{md2} = 1$	.65	x <sub>kd2</sub> =	1.76	R <sub>kd2</sub>	= .	0.0043
$X_{q2} = 1$	.73	$X_{kq2} =$	2.05	R <sub>kq2</sub>	=	0.0089
$X_{mg2} = 1$	.59	$R_{a2} =$	0.0015			

Mass-spring System

M <sub>HP2</sub>	=	0.248	17		01.0	
MTDO	=	0.464	<sup>K</sup> 12	=	21.0	
1Р <i>2.</i> М	_	2 31	<sup>к</sup> 23	Ξ	48.4	D = 0.1
LPA2		2.51	K.34	=	74.6	<sup>1</sup> ii <sup>0,1</sup>
MLPB2	=	2,38	К, -	=	62.3	i=1,2,,5
M Gen2	=	1.71	45			

Turbine Torque Distribution

F <sub>HP2</sub> =	0.3	$F_{IP2} =$	0.26	F <sub>LPA2</sub> =	0.22
F <sub>LPB2</sub> =	0.22		· · ·		

Exciter and Voltage Regulator

 $K_{A2} = 50$   $T_{A2} = 0.01$   $T_{E2} = 0.005$ 

Exciter voltage ceiling limits =  $\pm 7.0$ 

A2.3 Transmission System

Transformer

x <sub>t1</sub>	=	0.14	R <sub>t1</sub>	=	0.01
X <sub>t2</sub>	=	0.1	R <sub>t2</sub>	=	0.01

Transmission Line

Line 1	$\mathbf{x}_{L1}$	=	0.42		R <sub>L1</sub>	=	0.02	
	x <sub>c1</sub>	va	ried from	0.042	to O	.336	<b>(</b> 10%	- 80%)
Line 2	X <sub>L2</sub>	=	0.4		R <sub>L2</sub>	=	0.01	
	x <sub>c2</sub>	va	ried from	0.04	to 0.	32 (1	10% -	80%)
Line 3	X <sub>L3</sub>		0.2		R <sub>L3</sub>	=	0.01	
Linė 4	X L4	=	0.28		R <sub>L4</sub>	-	0.05	
Load					R <sub>10</sub>	ad	0.9	·

## APPENDIX III

### SYSTEM DATA FOR THE THREE MACHINE SYSTEM

#### A3.1 Machine 1

All parameters are similiar to those given in A2.1 of Appendix II. A3.2 Machine 2

All parameters are similiar to those given in A2.2 of Appendix II .

### A3.3 Machine 3

Synchronous Machine Parameters

$X_{d3} = 1.78$	$x_{f3} = 1.7781$	$R_{f3} = 0.00109$
$x_{md3} = 1.6845$	$x_{kd3} = 1.7368$	$R_{kd3} = 0.0117$
$x_{q3} = 1.7067$	$x_{kq3} = 1.6409$	$P_{kq3} = 0.0151$
$X_{mq3} = 1.6063$	$R_{a3} = 0.00357$	

Mass-spring System

$M_{\rm HP3} = 0.262$	K = 47.48	
$M_{IP3} = 0.523$	5 $12 $ $12 $ $12$	D - 01
M <sub>Gen3</sub> = 1.85	$\frac{1}{23} = 01.05$	ii <sup>D</sup> ii
$M_{Ex3} = 0.059$	$K_{34} = 4.51$	i=1,2,4

Turbine Torque Distribution

 $F_{HP3} = 0.4$   $F_{IP3} = 0.6$ 

Exciter and Voltage Regulator

 $K_{A3} = 50$   $T_{A3} = 0.02$ Exciter voltage ceiling limits =  $\pm 7.0$  129

 $T_{E3} = 0.002$ 

Transformer

X <sub>t1</sub>	=	0.14	R <sub>t1</sub>	` <b>=</b>	0.01
X <sub>t2</sub>	==	0.14	R <sub>t2</sub>	=	0.01
Xt3	=	0.14	R <sub>t3</sub>	=	0.01

# Transmission Line

Line 1	$X_{L1} = 0.47$	$R_{L1} = 0$	0.035
•	X varied from 30%	% to 70% c	compensation
Line 2	$X_{L2} = 0.4$	$R_{L2} = 0$	0.02
	X varied from 30%	% to 70% c	compensation
Line 3	$X_{L3} = 0.14$	$R_{1.3} = 0$	0.01
	X varied from 30%	% to 70% c	compensation
Line 4	$X_{L4} = 0.39$	$R_{j4} = 0$	0.03
	X <sub>c4</sub> varied from 30%	% to 70% c	compensation
Line 5	$X_{L5} = 0.34$	$R_{L5} = 0$	.025
	X varied from 30%	3 to 70% c	compensation
Line 6	$X_{L6} = 0.51$	$R_{L6} = 0$	0.04
	X varied from 30%	( to 70% c	compensation
Line 7	$X_{L7} = 0.11$	$R_{L7} = 0$	0.01
	No capacitor compens	sation	
Line 8	$X_{L8} = 0.3$	$P_{L8} = 0$	0.02
•	X <sub>c8</sub> varied from 30%	( to 70% c	compensation
Line 9	$X_{1.9} = 0.32$	$R_{L9} = 0$	.024